

ECSE 306 - Fall 2008

Fundamentals of Signals and Systems

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Lecture 24

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Bode Plots of Multi-Order (multi-pole-zero)
Systems

Frequency Responses of First-Order Lag, Lead
Systems

Steps of drawing Bode plots

Step 1: Dividing multi-order system into cascade of multiple 1st-order systems, each containing a single pole or zero;

Step 2: Determining the asymptotes and break frequencies of these 1st-order systems;

Step 3: Adding up the Bode plots of these 1st-order systems.

Example of a first-order system

Consider again the first-order system with frequency response

$$H(j\omega) = \frac{1}{j\omega + 2}.$$

It is convenient to write it as the **product** of a gain and a first-order transfer function **with unity gain at DC**:

$$H(j\omega) = \frac{1}{2} \frac{1}{j\omega/2 + 1}.$$

The Bode magnitude plot of the first-order system

The Bode plot of the magnitude is the graph of

$$\begin{aligned}20 \log_{10} |H(j\omega)| &= 20 \log_{10} \left| \frac{1}{2} \right| + 20 \log_{10} \left| \frac{1}{\frac{j\omega}{2} + 1} \right| dB \\ &= -20 \log_{10} 2 - 20 \log_{10} \left| \frac{j\omega}{2} + 1 \right| dB \\ &= -6 dB - 20 \log_{10} \left| \frac{j\omega}{2} + 1 \right| dB\end{aligned}$$

The Bode magnitude plot of a 1st-order system has **2 asymptotes**: **one straight line for very low frequencies**, and **one straight line for very high frequencies**. The frequency at which the two asymptotes meet is called the **break frequency**.

The low- and high-frequency asymptotes of the first-order system

For **low frequencies** ($\omega \ll 2$),

$$20 \log_{10} |H(j\omega)| \approx -6 \text{ dB} - 20 \log_{10} |1| \text{ dB} = -6 \text{ dB} .$$

i.e., for very low frequencies the Bode magnitude plot approximates a straight line.

For **high frequencies** ($\omega \gg 2$),

$$\begin{aligned} 20 \log_{10} |H(j\omega)| &\approx -6 \text{ dB} - 20 \log_{10} \left| \frac{\omega}{2} \right| \text{ dB} \\ &= -6 \text{ dB} - 20 \log_{10} |\omega| \text{ dB} + 20 \log_{10} 2 \text{ dB} \\ &= -20 \log_{10} |\omega| \text{ dB} \end{aligned}$$

i.e., for very high frequencies, the Bode magnitude plot approximates a straight line with a slope **-20 dB/decade**, or **-6 dB/octave**.

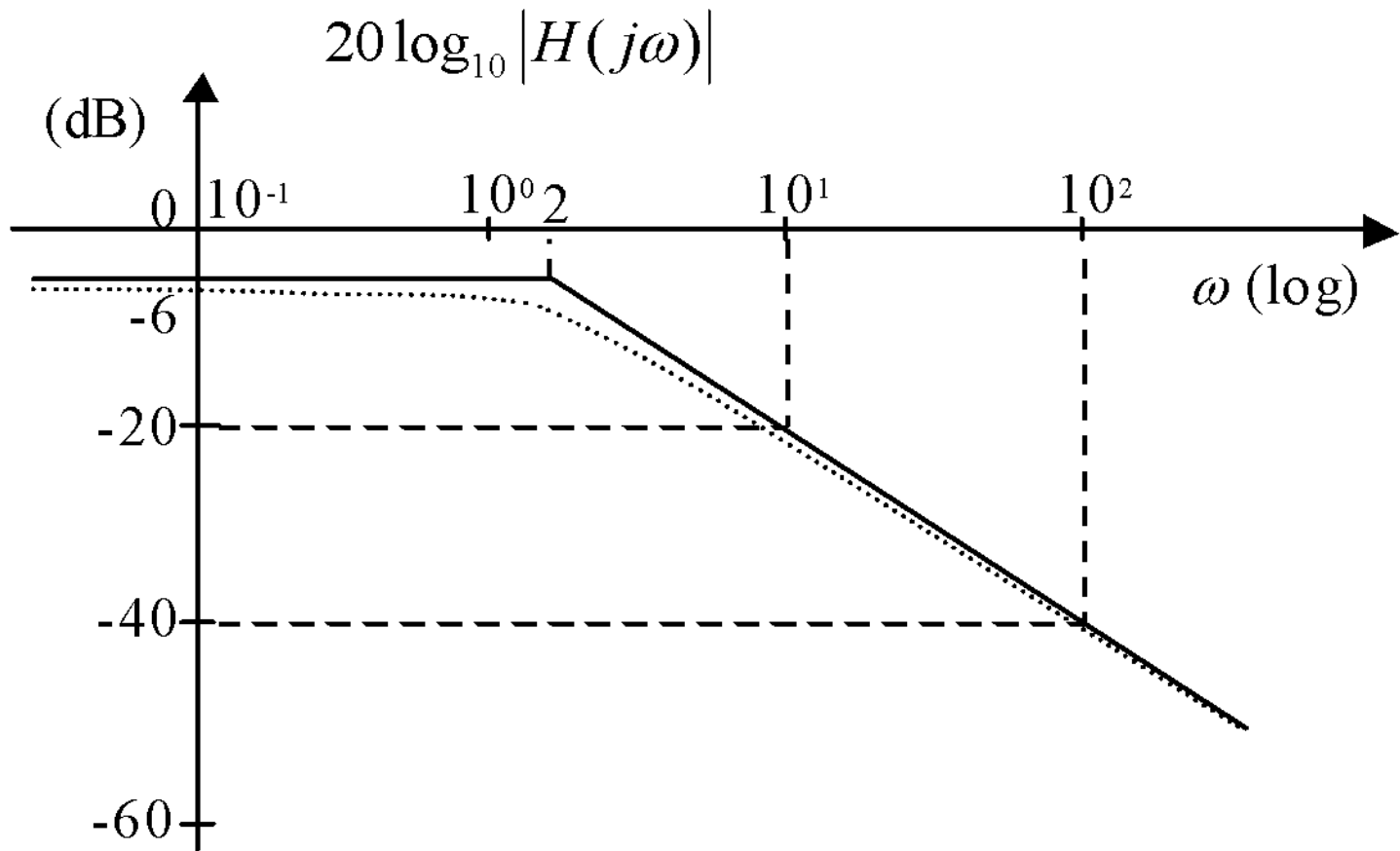
The slope of the high-frequency asymptotes

For $\omega \gg 2$, say $\omega = 10$, we get -20 dB; for $\omega = 100$, we get -40 dB, etc.

Therefore, the slope of the asymptote is -20dB/decade.

The 2 asymptotes meet at the break frequency 2 radians/s, at which the magnitude drops from the DC gain by 3 dB.

Given the 2 asymptotes and the drop at the break frequency, we can sketch the magnitude Bode plot (the dashed line as follows).



The asymptotes of the Bode phase plot of the 1st-order system

The phase response is given by:

$$\angle H(j\omega) = \angle \frac{1}{2} \frac{1}{j\omega/2 + 1} = \angle \frac{1}{j\omega/2 + 1} = \arctan\left(\frac{-\omega}{2}\right)$$

For $\omega \ll 2$, the phase approximates 0; for $\omega \gg 2$ the phase is

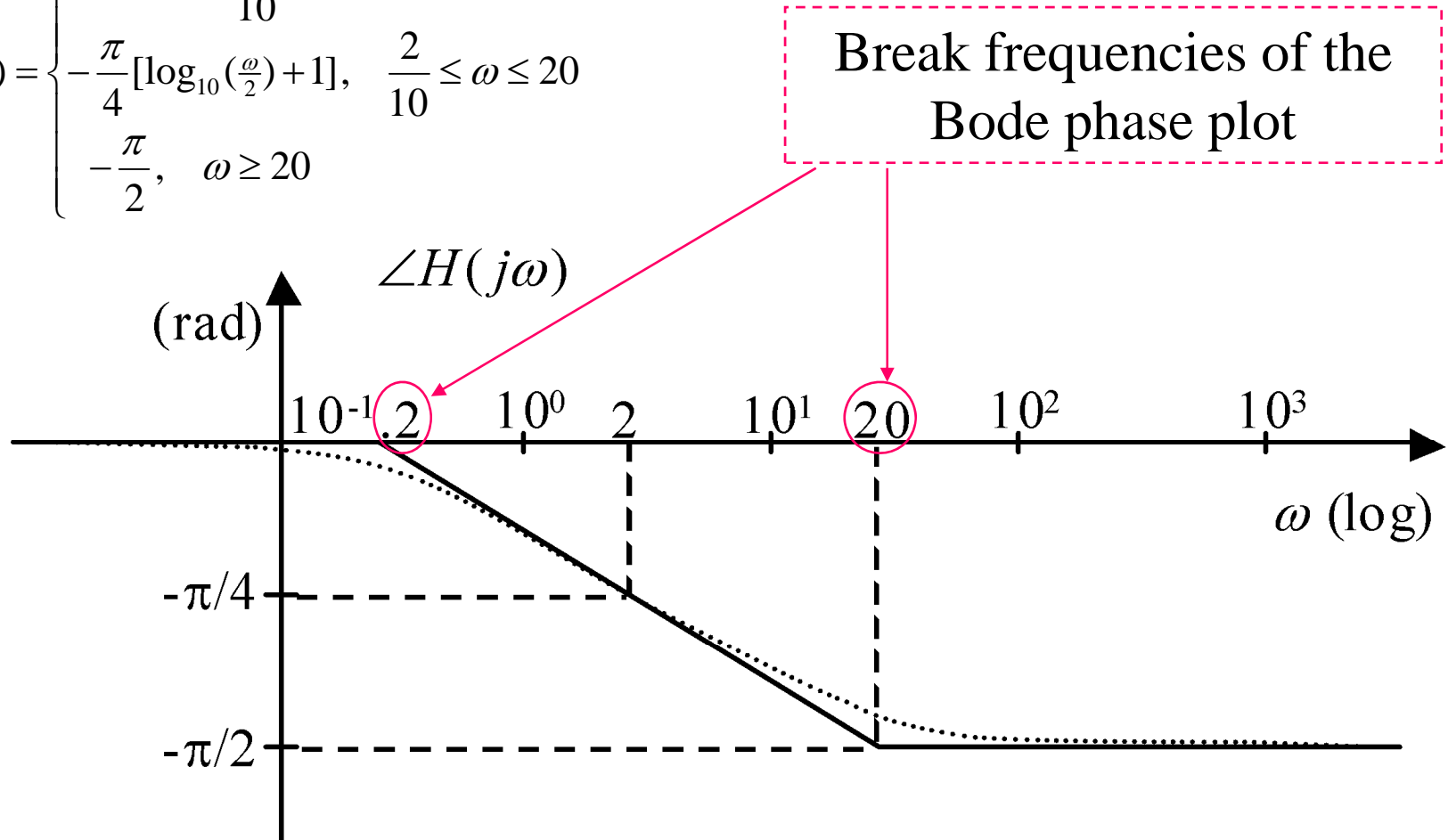
approximates $-\frac{\pi}{2}$. i.e., at very low and very high frequencies the phase response approximates to 2 parallel asymptotes, respectively

Connecting the 2 parallel asymptotes is a straight line, the third asymptotes. The three asymptotes are given by the following piecewise linear function:

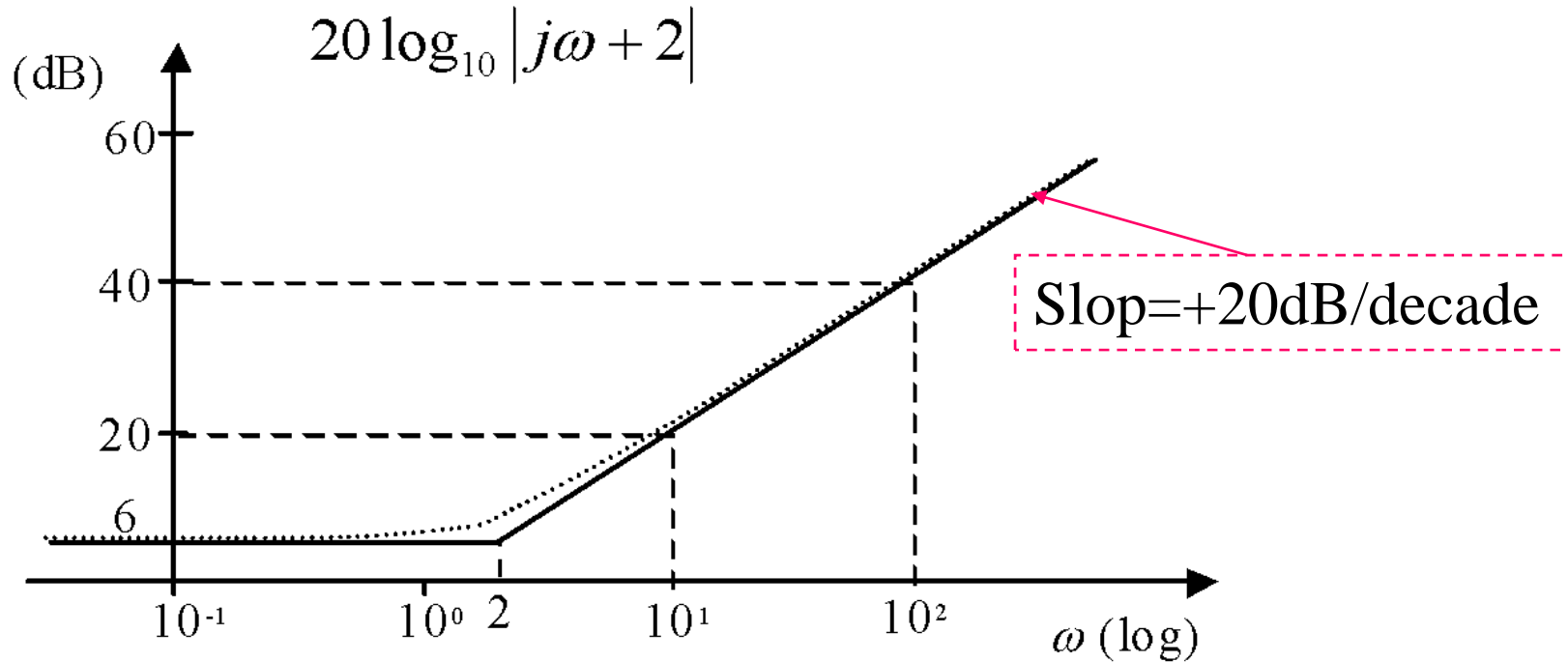
$$\angle H(j\omega) = \begin{cases} 0, & \omega \leq \frac{2}{10} \\ -\frac{\pi}{4} [\log_{10}(\frac{\omega}{2}) + 1], & \frac{2}{10} \leq \omega \leq 20 \\ -\frac{\pi}{2}, & \omega \geq 20 \end{cases}$$

The break frequencies of the Bode phase plot of the 1st-order system

$$\angle H(j\omega) = \begin{cases} 0, & \omega \leq \frac{2}{10} \\ -\frac{\pi}{4} [\log_{10}(\frac{\omega}{2}) + 1], & \frac{2}{10} \leq \omega \leq 20 \\ -\frac{\pi}{2}, & \omega \geq 20 \end{cases}$$



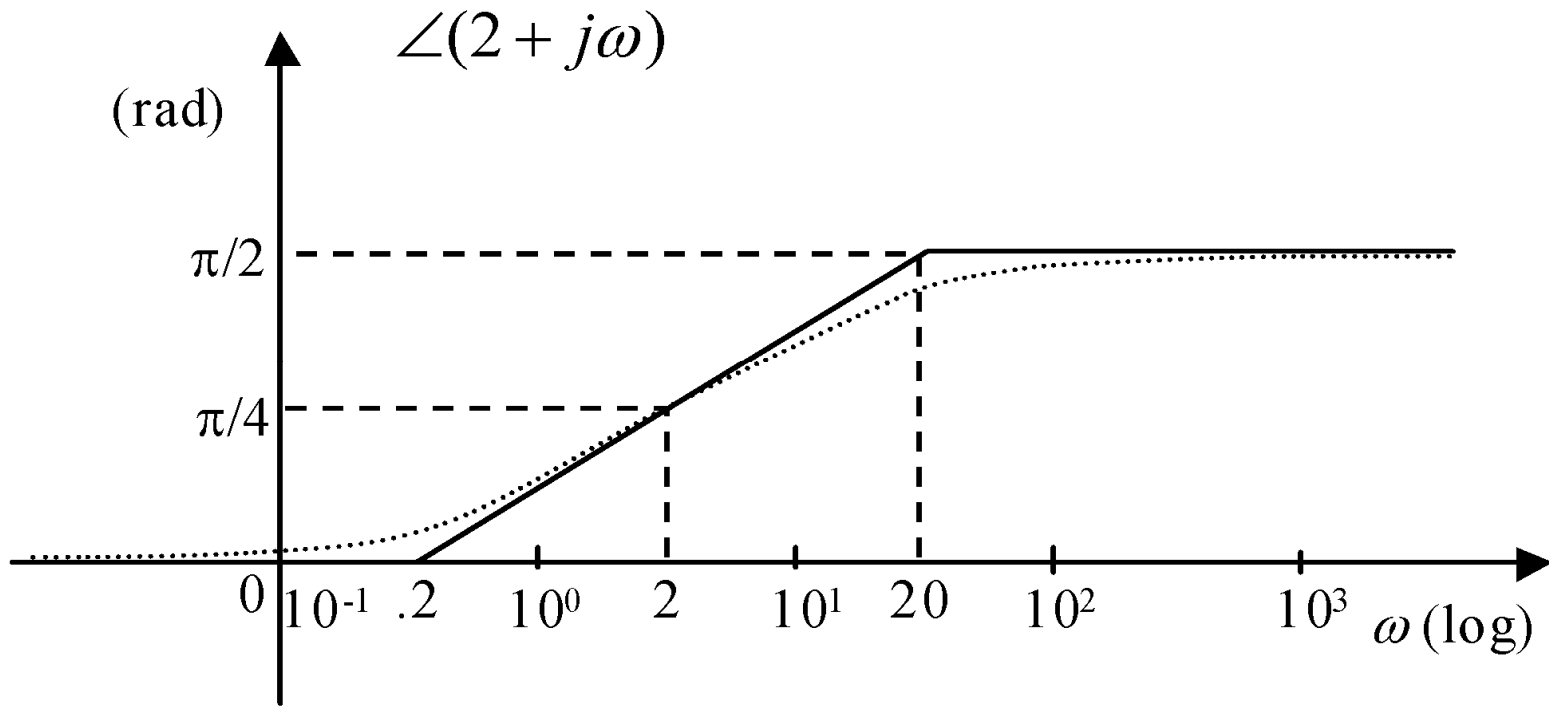
The Bode magnitude plot of a single-zero system



Given a numerator $(j\omega+2)$, i.e., the inverse of $1/(j\omega+2)$, the Bode magnitude plot is simply the Bode magnitude plot of $1/(j\omega+2)$ flipped around the frequency axis, because:

$$20 \log_{10} |j\omega + 2| = -20 \log_{10} \frac{1}{|j\omega + 2|}$$

The Bode phase plot of a single-zero system



Given a single-zero system $(s+2)$, i.e., the inverse of $1/(s+2)$, the Bode phase plot is simply the Bode phase plot of $1/(s+2)$ flipped around the frequency axis, because:

$$\angle(2 + j\omega) = -\angle \frac{1}{2 + j\omega}$$

The Bode plots of a second-order system

Second-Order Example:

$$\begin{aligned} H(s) &= \frac{s + 100}{(s^2 + 11s + 10)} = \frac{s + 100}{(s + 1)(s + 10)} \\ &= 10 \frac{\frac{s}{100} + 1}{(s + 1)\left(\frac{s}{10} + 1\right)}, \operatorname{Re}\{s\} > -1 \end{aligned}$$

which has the frequency response

$$H(j\omega) = 10 \frac{\frac{j\omega}{100} + 1}{(j\omega + 1)\left(\frac{j\omega}{10} + 1\right)}$$

The break frequencies are 1, 10 and 100 radians/s.

The Bode magnitude plot of the second-order system

The **Bode magnitude plot** is the graph of

$$\begin{aligned} 20\log_{10} |H(j\omega)| &= 20\log_{10} |10| + 20\log_{10} \left| \frac{j\omega}{100} + 1 \right| \\ &\quad + 20\log_{10} \left| \frac{1}{\frac{j\omega}{10} + 1} \right| + 20\log_{10} \left| \frac{1}{j\omega + 1} \right| \text{ dB} \\ &= 20 + 20\log_{10} \left| \frac{j\omega}{100} + 1 \right| - 20\log_{10} \left| \frac{j\omega}{10} + 1 \right| \\ &\quad - 20\log_{10} |j\omega + 1| \text{ dB} \end{aligned}$$

At low frequencies ($\omega \ll 1$),

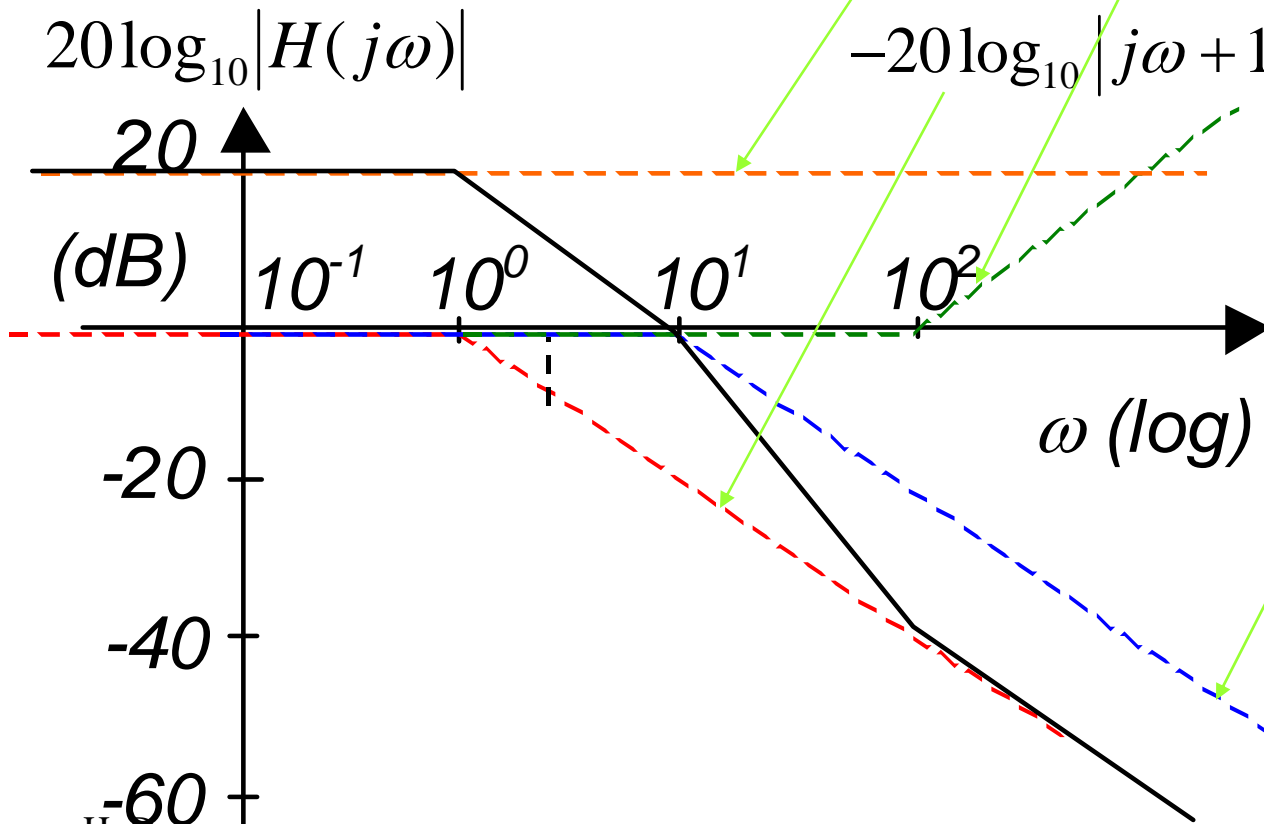
$$\begin{aligned} 20\log_{10} |H(j\omega)| &\approx 20 \text{ dB} + 20\log_{10} |1| - 20\log_{10} |1| \\ &\quad - 20\log_{10} |1| \text{ dB} = 20 \text{ dB} \end{aligned}$$

For **high frequencies** ($\omega \gg 100$),

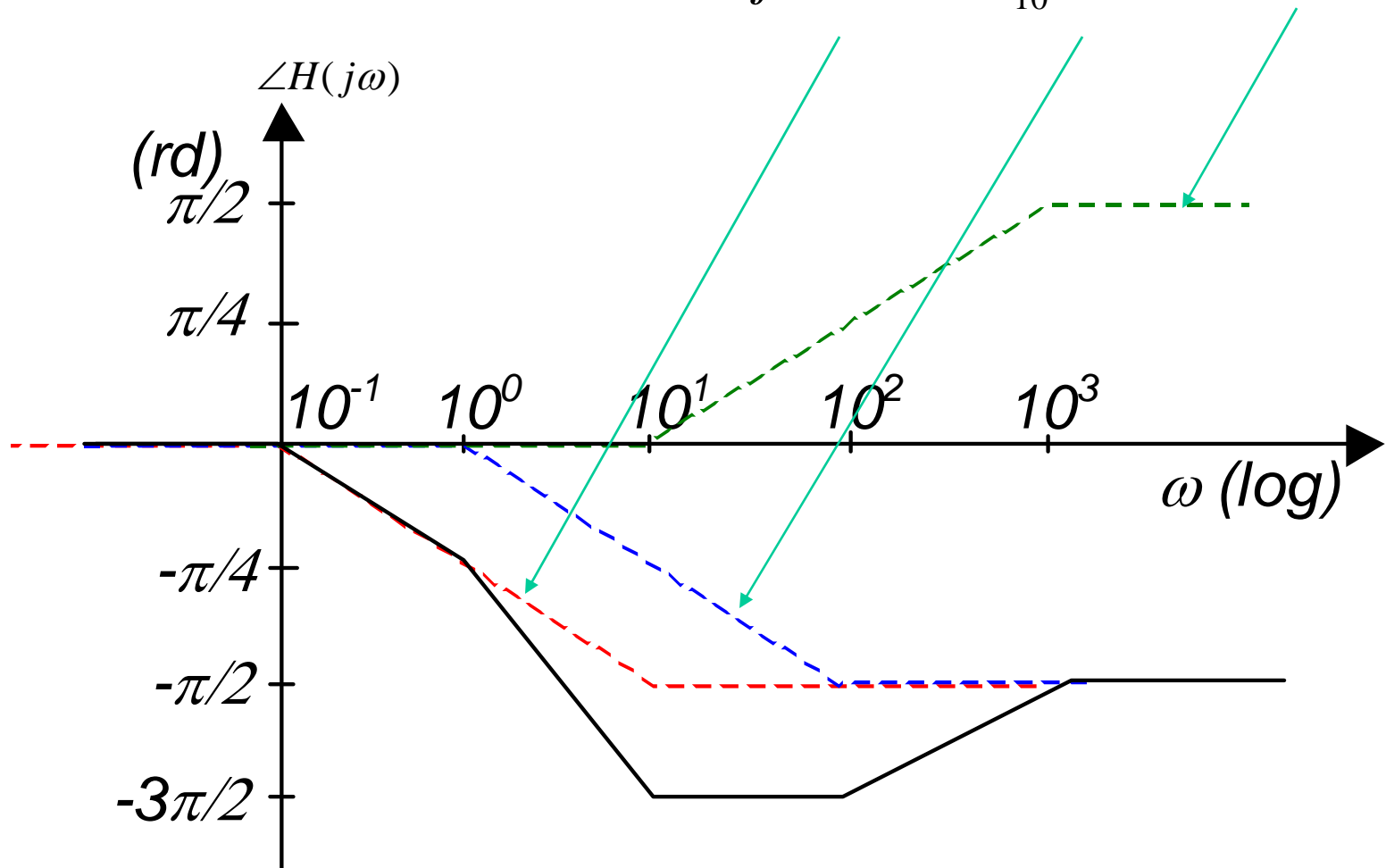
$$\begin{aligned} 20\log_{10}|H(j\omega)| &\approx 20 + 20\log_{10}\left|\frac{\omega}{100}\right| - 20\log_{10}\left|\frac{\omega}{10}\right| - 20\log_{10}|\omega| \text{ dB} \\ &= 20 + 20\log_{10}|\omega| - 40 - 20\log_{10}|\omega| + 20 - 20\log_{10}|\omega| \text{ dB} \\ &= -20\log_{10}|\omega| \text{ dB} \end{aligned}$$

We can plot the asymptotes of each first-order term on the same magnitude graph (dashed lines) and then add them together to obtain the Bode magnitude plot (solid line).

$$\begin{aligned}
 20 \log_{10} |H(j\omega)| &= 20 \log_{10} |10| + 20 \log_{10} \left| \frac{j\omega}{100} + 1 \right| \\
 &\quad + 20 \log_{10} \left| \frac{1}{\frac{j\omega}{10} + 1} \right| + 20 \log_{10} \left| \frac{1}{j\omega + 1} \right| \text{ dB} \\
 &= 20 + 20 \log_{10} \left| \frac{j\omega}{100} + 1 \right| - 20 \log_{10} \left| \frac{j\omega}{10} + 1 \right| \\
 &\quad - 20 \log_{10} |j\omega + 1| \text{ dB}
 \end{aligned}$$



$$\angle H(j\omega) = \angle \frac{1}{j\omega + 1} + \angle \frac{1}{\frac{j\omega}{10} + 1} + \angle (j\omega + 1)$$



First-order lag system

A **first-order lag** has a transfer function of the form

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1} ,$$

where $0 \leq \alpha < 1$, $\tau > 0$ is the time constant.

This system is called a *lag* because it has **an effect similar to a pure delay** $e^{-\tau s}$ **at low frequencies** for $\alpha = 0$. To the first order, the two systems are the same, as can be seen from the Taylor series around $s=0$:

$$\frac{1}{\tau s + 1} = 1 - \tau s + (\tau s)^2 - (\tau s)^3 + \dots \approx 1 - \tau s$$

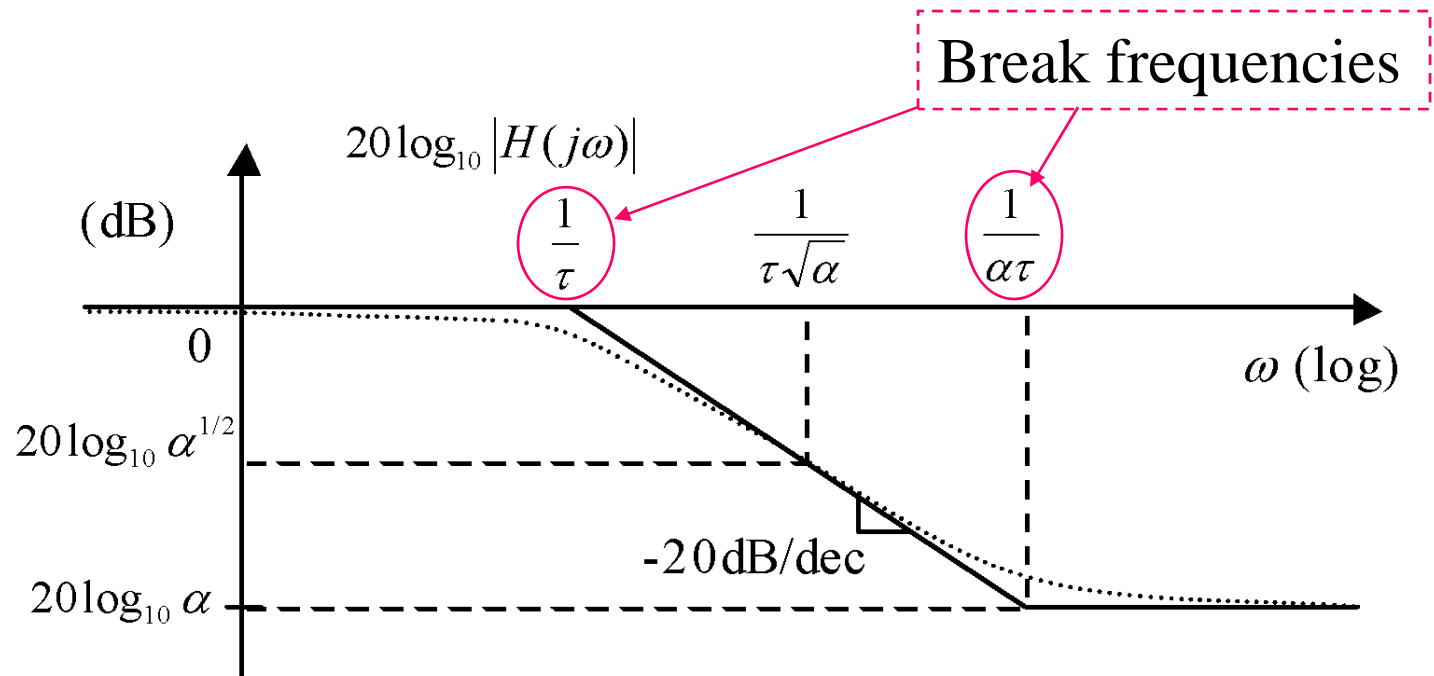
$$e^{-\tau s} = 1 - \tau s + \frac{1}{2}(\tau s)^2 - \frac{1}{3!}(\tau s)^3 + \dots \approx 1 - \tau s$$

The Bode magnitude plot of the first-order lag system

Assuming that $0 < \alpha < 1$, the frequency response of the first-order lag system is

$$H(j\omega) = \frac{j\omega\alpha\tau + 1}{j\omega\tau + 1}$$

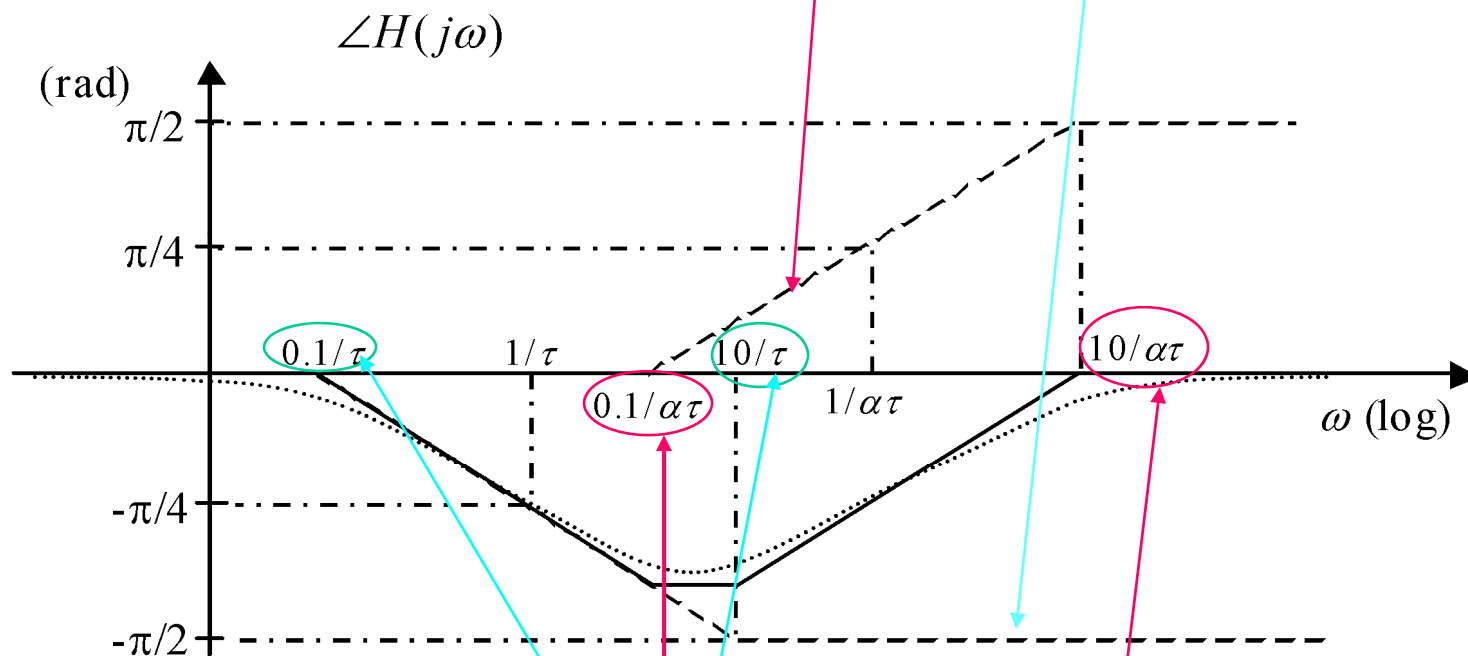
The Bode *magnitude* plot of the above system is shown below.



The phase Bode plot of the first-order lag system

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1} \quad \text{Re}\{s\} > -\frac{1}{\tau}$$

$$\angle H(j\omega) = \angle \frac{j\omega\alpha\tau + 1}{j\omega\tau + 1} = \angle(j\omega\alpha\tau + 1) + \angle \frac{1}{j\omega\tau + 1}$$



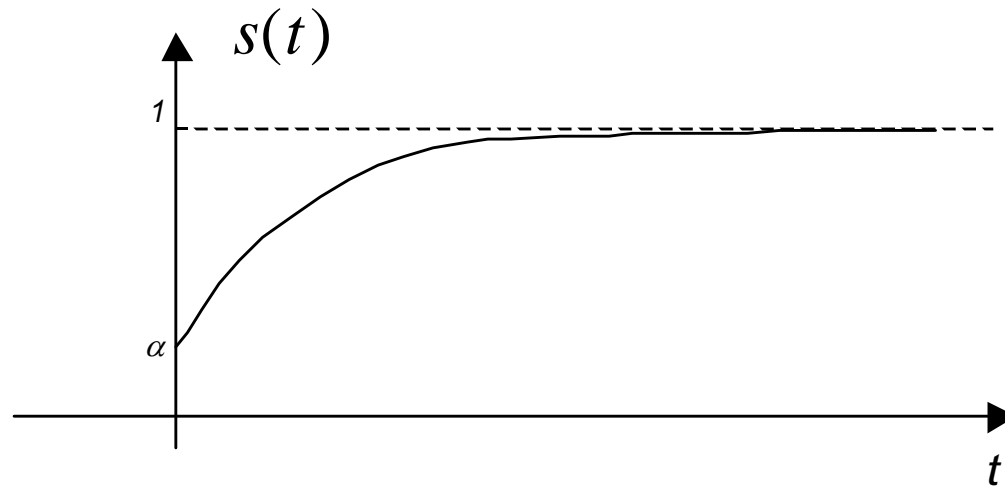
The step response of the first-order lag system

For the case $0 < \alpha < 1$ (lag) and time constant τ , i.e.,

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1} = \alpha + \frac{1 - \alpha}{\tau s + 1},$$

the step response is

$$s(t) = \alpha u(t) + (1 - \alpha)(1 - e^{-\frac{t}{\tau}})u(t).$$



First-Order Lead

The transfer function of a **first-order lead** system has the same algebraic expression as that of first-order lag system:

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1}$$

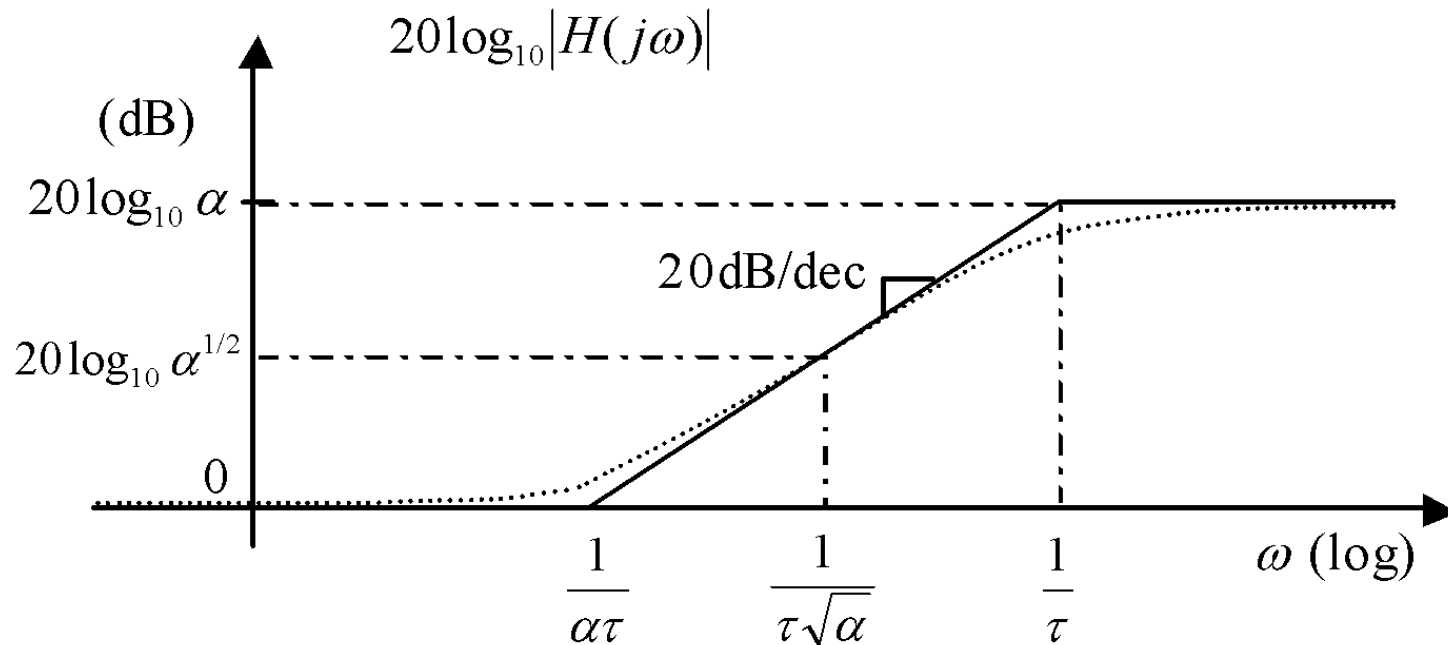
where $\tau > 0$, but $\alpha > 1$.

For $\alpha > 1$, the break frequency ($1/\alpha\tau$) given by the numerator is smaller than that ($1/\tau$) given by the denominator.

Recall: in a first-order lag system, $\alpha < 1$, and the break frequency ($1/\alpha\tau$) given by the numerator is greater than that ($1/\tau$) given by the denominator.

The Bode magnitude plot

For a first-order lead system, in which $\alpha > 1$, the magnitude increase from a level to a higher level as frequency increases.

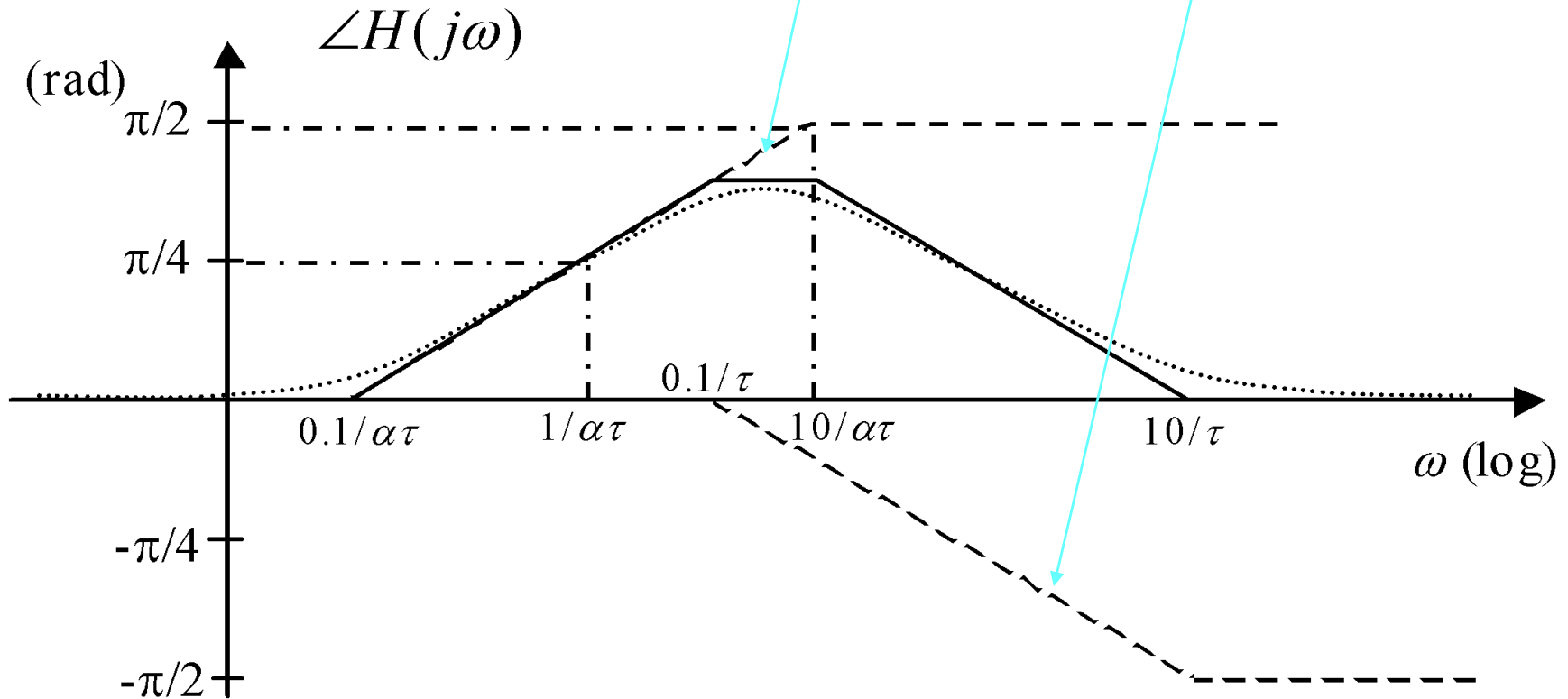


Compare with the first-order lag system, in which $\alpha < 1$, the magnitude decrease from a level to a lower level for a system.

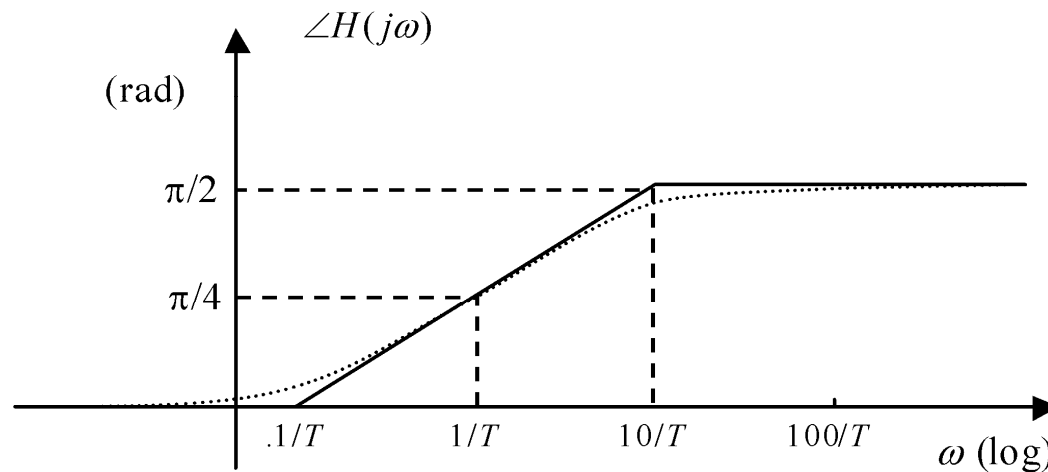
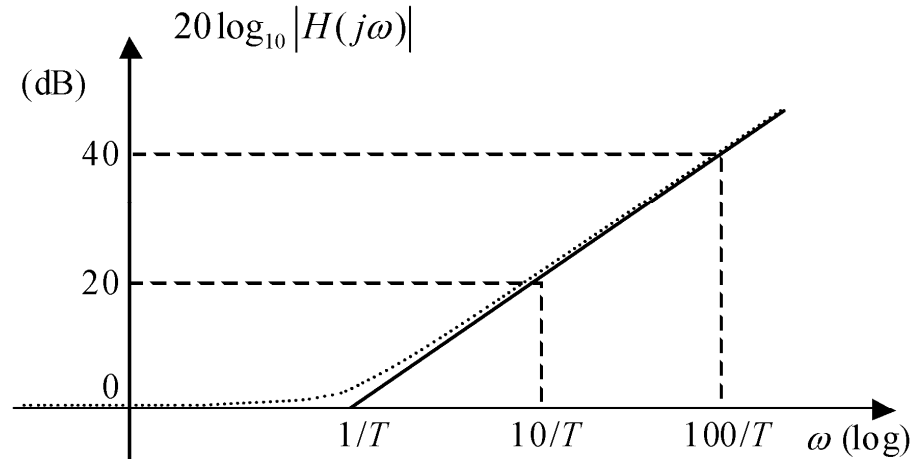
The Bode phase plot

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1} \quad \text{Re}\{s\} > -\frac{1}{\tau}$$

$$\angle H(j\omega) = \angle \frac{j\omega\alpha\tau + 1}{j\omega\tau + 1} = \angle(j\omega\alpha\tau + 1) + \angle \frac{1}{j\omega\tau + 1}$$



For the case where $\tau \rightarrow 0$, $\alpha\tau \rightarrow T$, then the first-order lead is equivalent to a differentiator with gain T in parallel with the identity system: $H(s) = Ts + 1$



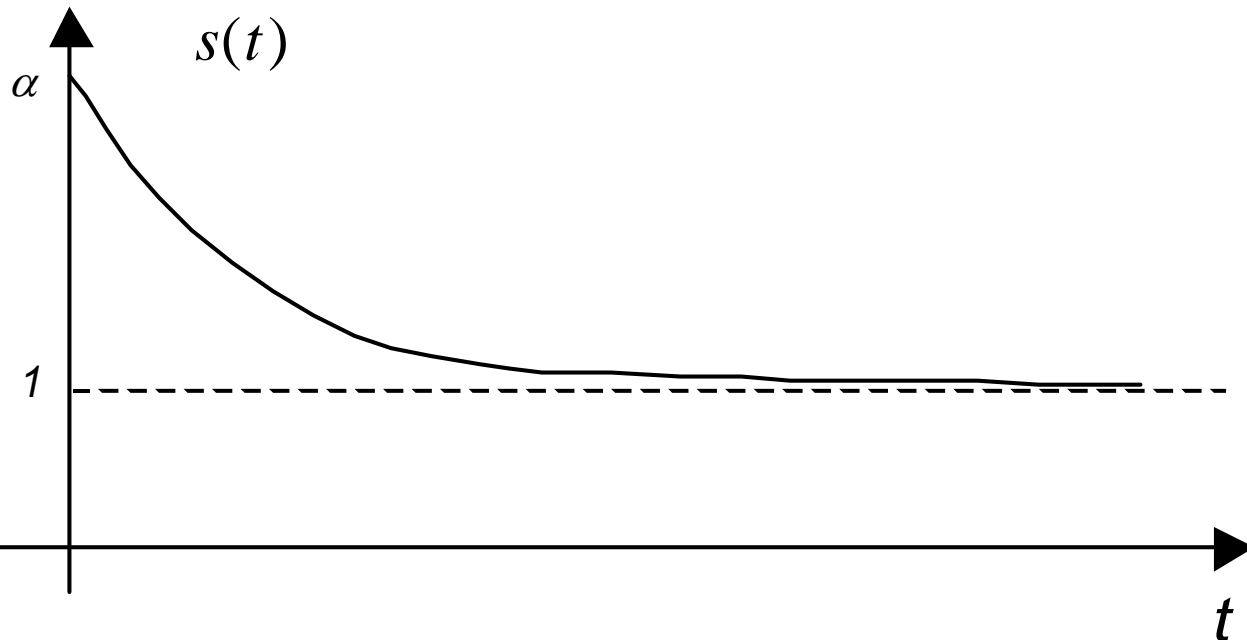
The step response of the first-order lead system

For the case $\alpha > 1$ (lead) and time constant τ , i.e.,

$$H(s) = \frac{\alpha\tau s + 1}{\tau s + 1} = \alpha + \frac{1 - \alpha}{\tau s + 1},$$

the step response is

$$s(t) = \alpha u(t) + (1 - \alpha)(1 - e^{-\frac{t}{\tau}})u(t).$$



Applications of first-order lead systems

The first-order lead may be used

- To "differentiate" signals at frequencies higher than $(\alpha\tau)^{-1}$ but lower than τ^{-1} .
- To "reshape" pulses that could have been distorted by a communication channel with a lowpass frequency response.
- As a controller because it adds positive phase to the overall loop transfer function .