

Solutions to Assignment 10

10.1 Exercise 13.6 of Boulet's book;

Compute the inverse z -transform of $X(z) = \frac{z^2}{z+0.2}$, $0.2 < |z| < \infty$ using the power series expansion method.

Answer:

$$X(z) = \frac{z^1}{(1+0.2z^{-1})} = \frac{z^2}{z+0.2}, \quad 0.2 < |z| < \infty$$

Long division yields:

$$\begin{array}{r} z - 0.2 + 0.04z^{-1} - 0.008z^{-2} \dots \\ 1 + 0.2z^{-1} \overline{)z} \\ \underline{z + 0.2} \\ -0.2 \\ \underline{-0.2 - 0.04z^{-1}} \\ 0.04z^{-1} \\ \underline{0.04z^{-1} + 0.008z^{-2}} \\ -0.008z^{-2} \end{array}$$

Note that the resulting power series converges because the ROC implies $|0.2z^{-1}| < 1$. The signal

is:

$$\begin{aligned} x[n] &= \delta[n+1] - 0.2\delta[n] + (0.2)^2\delta[n-1] - (0.2)^3\delta[n-2] + \dots \\ &= (-0.2)^{n+1}u[n+1] \end{aligned}$$

10.2 Exercise 13.8 of Boulet's book.

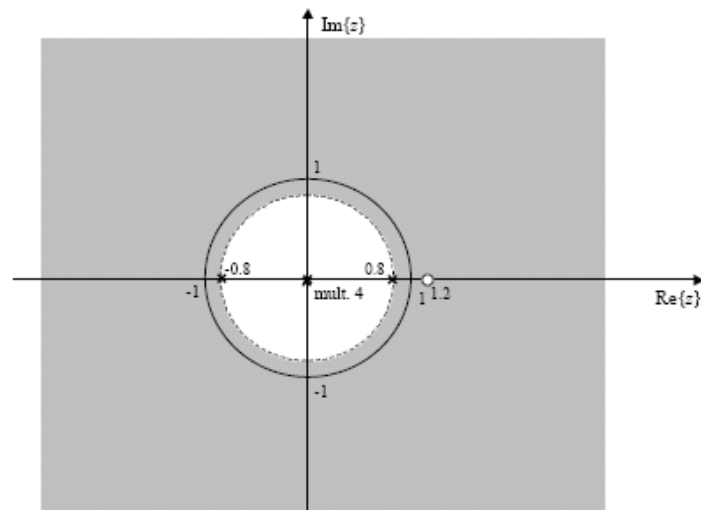
Consider a DLTI system with transfer function $H(z) = \frac{z^{-3} - 1.2z^{-4}}{(z - 0.8)(z + 0.8)}$.

(a) Sketch the pole-zero plot of the system.

Answer:

$$H(z) = \frac{z^{-5}(1 - 1.2z^{-1})}{(1 - 0.8z^{-1})(1 + 0.8z^{-1})} = \frac{z - 1.2}{z^4(z - 0.8)(z + 0.8)}$$

poles at -0.8 , 0.8 and 0 (multiplicity 4), zeros at 1.2 and at infinity.



(b) Find the ROC that makes this system stable.

Answer:

For stability, the ROC must include the unit circle. Hence, the ROC is $|z| > 0.8$, as shown on the above pole-zero plot.

(c) Is the system causal with the ROC that you found in (b)? Justify your answer.

Answer:

For the system to be causal, we must have $\lim_{z \rightarrow \infty} H(z) < \infty$, which is the case here ($=0$). Hence, the system is causal.

(d) Suppose that $H(e^{j\omega})$ is bounded for all frequencies. Find the response of the system $y[n]$ to the input $x[n] = u[n]$.

Answer:

This means that the ROC is chosen to be $|z| > 0.8$. We have $X(z) = \frac{1}{1-z^{-1}}$, $|z| > 1$, and

$$Y(z) = H(z)U(z) = \frac{z^{-5}(1-1.2z^{-1})}{(1-0.8z^{-1})(1+0.8z^{-1})(1-z^{-1})}, |z| > 1$$

We will treat the time delay later, so let's do a partial fraction expansion of:

$$\frac{(1-1.2z^{-1})}{(1-0.8z^{-1})(1+0.8z^{-1})(1-z^{-1})} = \frac{-0.56}{1-z^{-1}} + \frac{1}{1-0.8z^{-1}} + \frac{0.56}{1+0.8z^{-1}}$$

where the ROC of each term is taken to be the exterior of a circle of radius equal to the magnitude of the pole. This gives:

$$w[n] = (0.8)^n u[n] + 0.56(-0.8)^n u[n] - 0.56u[n]$$

and adding the time delay yields:

$$y[n] = (0.8)^{n-5} u[n-5] + 0.56(-0.8)^{n-5} u[n-5] - 0.56u[n-5]$$

10.3 Exercise 13.10 of Boulet's book

Also, for the $H(z)$ in this problem, please determine the corresponding difference equation relating the output $y[n]$ to the input $x[n]$.

Sketch the pole-zero plot and compute the impulse response $h[n]$ of the stable system with

$$\text{transfer function: } H(z) = \frac{2000z^3 + 1450z^2 + 135z}{(100z^2 - 81)(5z + 4)}.$$

Specify its ROC. Specify whether the system is causal or not.

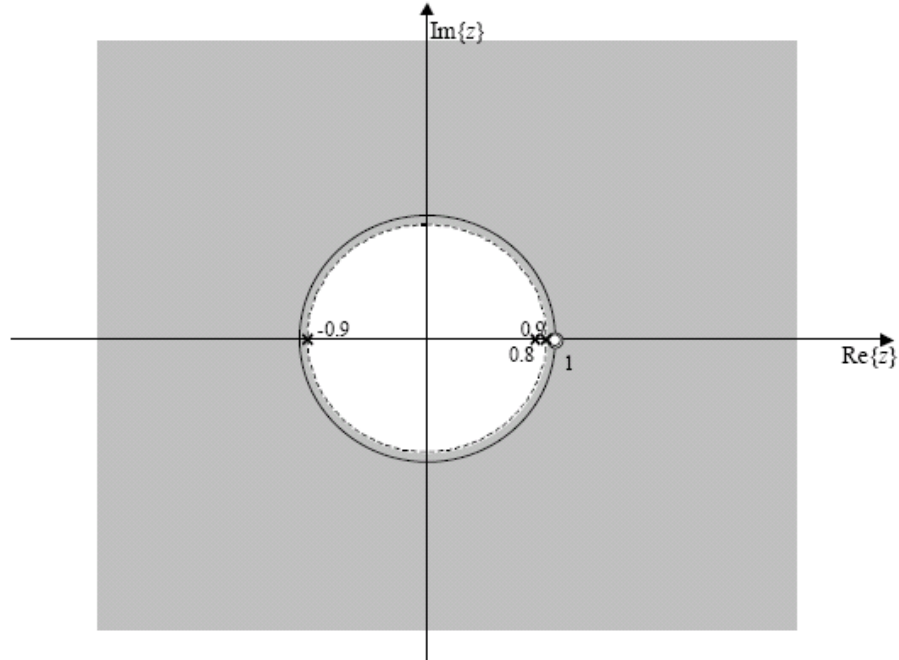
Answer:

$$\begin{aligned} H(z) &= \frac{2000z^3 + 1450z^2 + 135z}{(100z^2 - 81)(5z + 4)}, \\ &= \frac{4 + 2.9z^{-1} + 0.27z^{-2}}{(1 + 0.8z^{-1})(1 + 0.9z^{-1})(1 - 0.9z^{-1})}, \quad |z| > 0.9 \\ &= \left(\underbrace{\frac{-3}{(1 + 0.8z^{-1})}}_{|z| > 0.8} + \underbrace{\frac{5}{(1 + 0.9z^{-1})}}_{|z| > 0.9} + \underbrace{\frac{2}{(1 - 0.9z^{-1})}}_{|z| > 0.9} \right) \end{aligned}$$

The inverse z -transform is obtained using Table D.10:

$$h[n] = -3(-0.8)^n u[n] + 2(0.9)^n u[n] + 5(-0.9)^n u[n].$$

The pole-zero plot is shown below.



The system is causal since ROC is the outside of a disk, including infinity.

For the $H(z)$, the corresponding difference equation can be derived according to:

$$\frac{Y(z)}{X(z)} = \frac{2000 + 1450z^{-1} + 135z^{-2}}{500 + 400z^{-1} - 405z^{-2} - 324z^{-3}}$$

$$Y(z)(500 + 400z^{-1} - 405z^{-2} - 324z^{-3}) = X(z)(2000 + 1450z^{-1} + 135z^{-2})$$

Thus the difference equation is:

$$500y[n] + 400y[n-1] - 405y[n-2] - 324y[n-3] = 2000x[n] + 1450x[n-1] + 135x[n-2]$$

10.4 See Eq. (14.50) of Boulet's book. Now, the impulse response of a DT filter is defined by

$$h[n] = \begin{cases} 1/10, & n = 0, 1, \dots, 9 \\ 0, & \text{otherwise} \end{cases}$$

- Determine the transfer function of the system.
- Determine the frequency response of the filter.
- Sketch the poles and zeros of the system in the z -plane.

Answer:

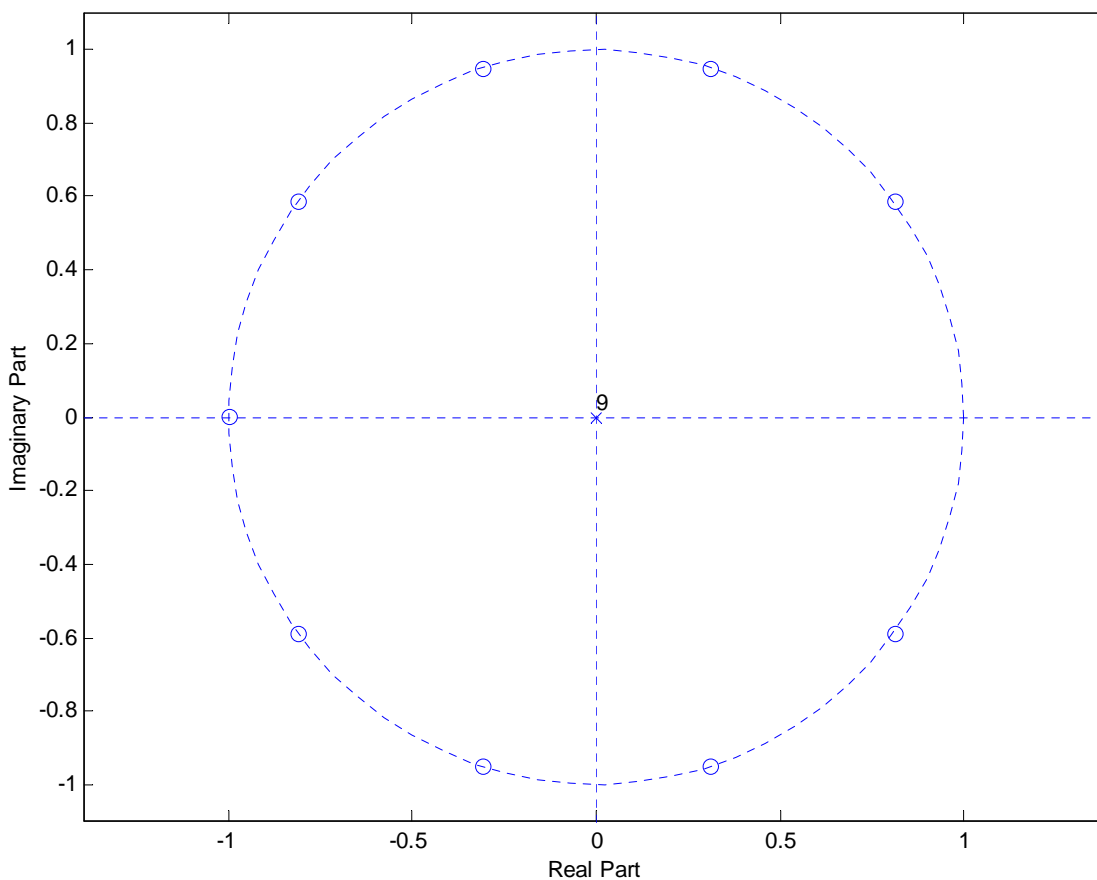
- Its transfer function can be obtained from the z -transform of the impulse response:

$$H(z) = \sum_{n=0}^9 h[n]z^{-n} = \frac{1}{10} [1 + z^{-1} + \dots + z^{-9}] = \frac{1}{10} \frac{1 - z^{-10}}{1 - z^{-1}} = \frac{1}{10} \frac{z^{10} - 1}{z^9(z-1)}$$

(b) The frequency response of the filter is

$$H(z = e^{j\omega}) = \frac{1}{10} \frac{1 - e^{-j10\omega}}{1 - e^{-j\omega}} = \frac{1}{10} \frac{e^{-j5\omega} (e^{j5\omega} - e^{-j5\omega})}{e^{-j\omega/2} (e^{j\omega/2} - e^{-j\omega/2})} = \frac{1}{10} e^{-j9\omega/2} \frac{\sin(5\omega)}{\sin(\omega/2)}$$

(c) The pole at $z=1$ is cancelled by the zero at $z=1$. Thus, there are 9 poles at $z=0$, and 9 zeros at $e^{j\frac{2\pi}{10}k}$, $k = 1, 2, \dots, 9$ in the z -plane, as shown in the figure below.



10.5 Find the inverse z-transform of

$$X(z) = \frac{1 - \frac{1}{1024} z^{-10}}{(1 - \frac{1}{2} z^{-1})}, \quad |z| > 0$$

Answer:

Since the ROC includes the entire z -plane, we know that the signal must be finite length. From the finite sum formula, we have

$$\left[\frac{1 - \frac{1}{1024} z^{-10}}{1 - \frac{1}{2} z^{-1}} \right] = \sum_{n=0}^9 \left(\frac{1}{2}\right)^n z^{-n}$$

Comparing this with the definition of z-transform, we have

$$x[n] = \begin{cases} (\frac{1}{2})^n, & 0 \leq n \leq 9 \\ 0, & \text{otherwise} \end{cases}$$

10.6 Consider a system whose input $x[n]$ and output $y[n]$ are related by

$$y[n-1] + 2y[n] = x[n]$$

- Determine the zero-input response of the system if $y[-1]=2$.
- Determine the zero-state response of the system to the input $x[n]=(1/4)^n u[n]$.
- Determine the output of the system for $n \geq 0$ when the initial condition and the input are given as in (a) and (b).

Answer:

Applying the unilateral z-transform to the difference equation, we have:

$$z^{-1}\mathcal{Y}[z] + y[-1] + 2\mathcal{Y}[z] = \mathcal{X}[z]$$

- For the zero-input response, assume that $x[n]=0$. Since we have given that $y[-1]=2$, we have

$$z^{-1}\mathcal{Y}[z] + y[-1] + 2\mathcal{Y}[z] = 0$$

$$\mathcal{Y}[z] = \frac{-2}{2 + z^{-1}} = \frac{-1}{1 + \frac{1}{2}z^{-1}}$$

Taking the inverse z-transform, we have the zero-input response:

$$y[n] = -(-\frac{1}{2})^n u[n]$$

- For the zero-state response, set $y[-1]=0$. Also we have

$$\mathcal{X}(z) = \frac{1}{1 - \frac{1}{4}z^{-1}}, \quad |z| > 1/4$$

Therefore,

$$\mathcal{Y}[z] = \left(\frac{1}{1 - \frac{1}{4}z^{-1}}\right) \left(\frac{1/2}{1 + \frac{1}{2}z^{-1}}\right)$$

By doing partial fraction expansion, we get:

$$\mathcal{Y}[z] = \frac{1/6}{1 - \frac{1}{4}z^{-1}} + \frac{1/3}{1 + \frac{1}{2}z^{-1}}$$

Taking inverse z-transform, we get the zero-state response:

$$y[n] = \frac{1}{6} \left(\frac{1}{4}\right)^n u[n] + \frac{1}{3} \left(-\frac{1}{2}\right)^n u[n]$$

- The total response is the sum of zero-input response and the zero-state response:

$$y[n] = \frac{1}{6} \left(\frac{1}{4}\right)^n u[n] - \frac{2}{3} \left(-\frac{1}{2}\right)^n u[n]$$

10.7 Consider a signal $x(t)=\cos(200\pi t)$.

(a) Determine the sampling frequency F_s such that $x(t)$ can be recovered via low-pass filtering of the sampled signal $x(nT_s)$, where $T_s=1/F_s$.

(b) If $F_s=120$ Hz, what frequency components are contained in the sampled signal $x(nT_s)$?

Answer:

(a) The frequency of the signal $x(t)$ is 100 Hz.

According to the sampling theory, the sampling frequency F_s must be greater than $2f_M$, where $f_M=100$ Hz. Thus, F_s must be greater than 200 Hz.

(b) If $F_s=120$ Hz, then the signal $x[nT_s]$ contain frequency components at $\pm f_M+kF_s$, $k=\dots,-3, -2, -1, 0, 1, 2, 3, \dots$

Considering positive frequencies, we know the frequency components are located at:

$$100+120=220 \text{ Hz,}$$

$$-100+120=20 \text{ Hz,}$$

$$100+240=340 \text{ Hz,}$$

$$-100+240 =140 \text{ Hz,}$$

$$100+360=460 \text{ Hz,}$$

$$-100+360=260 \text{ Hz,}$$

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