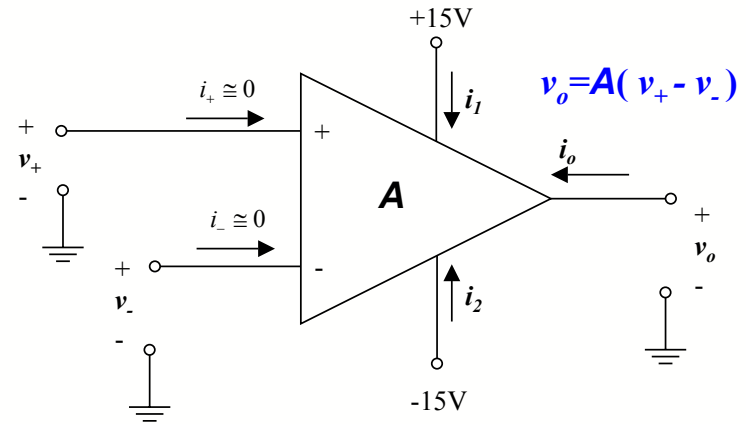


## ECSE 210: Circuit Analysis

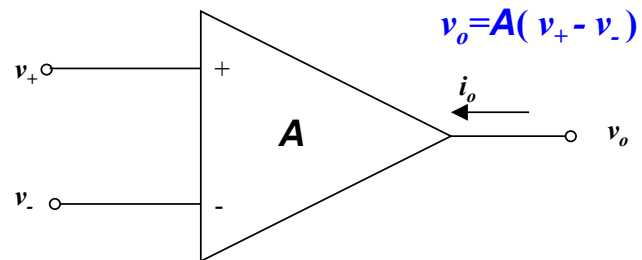
### Lecture #4: Operational Amplifiers

## OpAmp Symbol



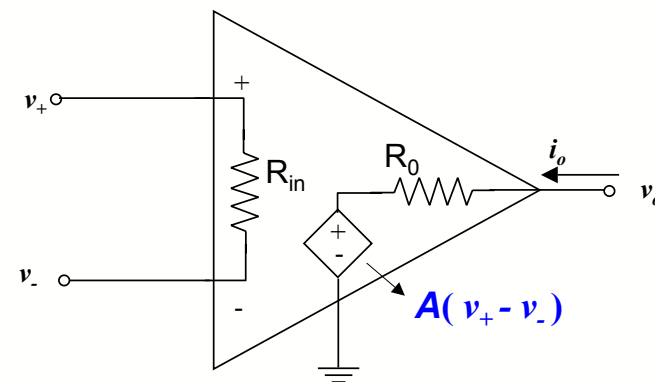
➔ A **differential** amplifier.

## Simplified OpAmp Symbol

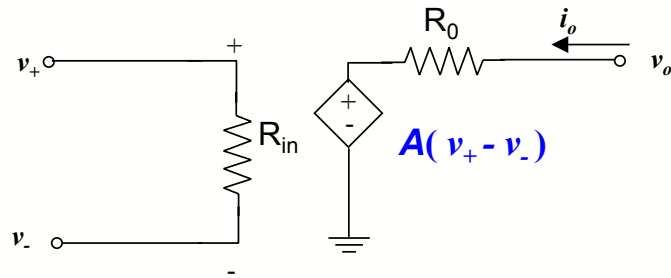


1. Supply voltages are not shown.
2. Supply currents are not shown. (Be careful with KCL)
3. Never do KCL at the reference node (otherwise need to include supply currents).
4. KCL at output node  $v_o$  is never done unless we want to calculate  $i_o$ .

## OpAmp Model

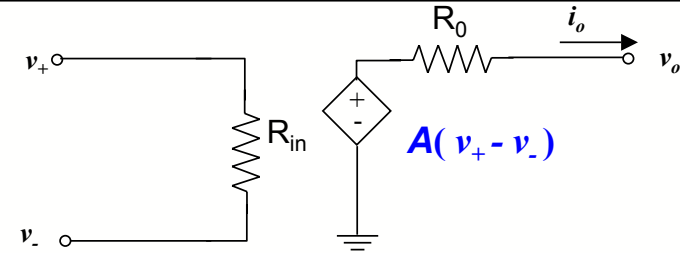


## Op-Amp Equivalent Circuit



- ➔ Based on this circuit, the op-amp is a unilateral device.
- The output voltage is determined as a function of the input, **BUT the input voltages are not affected by the output voltage.**

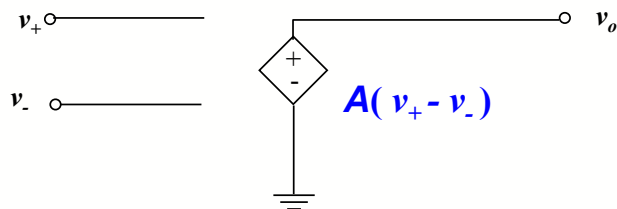
## Some Practical Information



1. The input resistance  $R_{in}$  is very large, typically in the range of  $100\text{k}\Omega \rightarrow 1\text{T}\Omega$ .
2. The voltage gain  $A$  is very high, typically  $10^5 \rightarrow 10^7$ .
3. The output resistance  $R_o$  is very small **compared** to the recommended output load. Typically  $R_o$  is in the range  $1\Omega \rightarrow 75\Omega$ .

➔ These parameters suggest a simpler model!

## Simpler Op-Amp Model



1. Since  $R_{in}$  is very large, assume  $R_{in} = \text{infinity}$ .
2. Since  $R_o$  is very small assume  $R_o = \text{zero}$
3. This model appears on page 84 of the text, Figure 3.7. It depends only on the **open loop** gain  $A$ .

## The Ideal OpAmp Model

Assume the “OpAmp Equivalent Circuit” :

1. Since  $R_{in}$  is very large, assume  $R_{in} = \text{infinity}$ .
2. Since  $R_o$  is very small assume  $R_o = \text{zero}$ .
3. Since  $A$  is very large, assume  $A = \text{infinity}$ .

} Same as previous slide

This gives the “**Ideal OpAmp Model**”

## Condition of Linearity

- The *ideal* op-amp is linear.
- For an op-amp to be linear:

$$|v_o| \leq v_{sat} \quad |i_o| \leq i_{sat} \quad \left| \frac{dv_o}{dt} \right| \leq SR$$

- What is  $v_{sat}$ ?
- What is SR (slew rate)?

## Op-Amp Properties

### → Good:

- High input impedance.
- Low output impedance.
- Large gain.

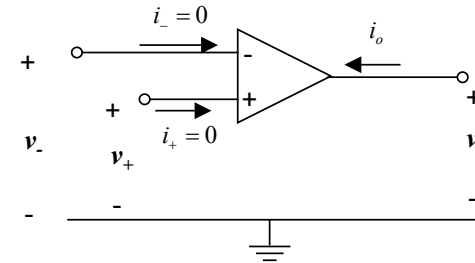
### → Bad:

The properties of the op-amp (such as its gain A) are strongly dependent on process variations, temperature variations, etc.

→ Use closed loop, **negative feedback** configuration.

## Virtual Short / Virtual Open

→ Virtual short / virtual open principles (*Textbook p.151*).

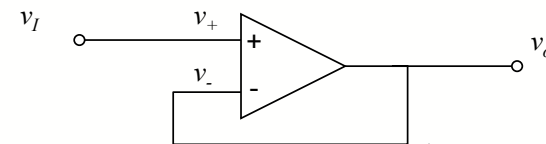


- Virtual open:  $i_- = 0, i_+ = 0$
- Virtual short:  $v_+ = v_-$

→ Note:  $i_o$  is not equal to zero!

→ Op-Amp must function in the linear region.

## Voltage Follower



Negative Feedback

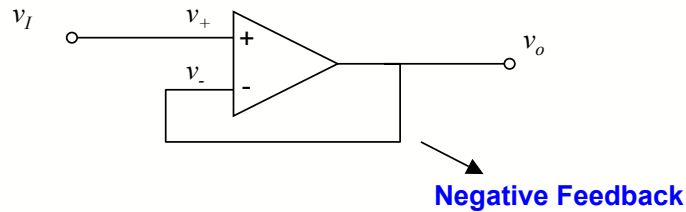
$$v_o = A(v_+ - v_-)$$

$$v_+ = v_I \quad v_- = v_o$$

$$v_o = A(v_I - v_o) = Av_I - Av_o$$

$$\frac{v_o}{v_I} = \frac{A}{A+1} \cong 1$$

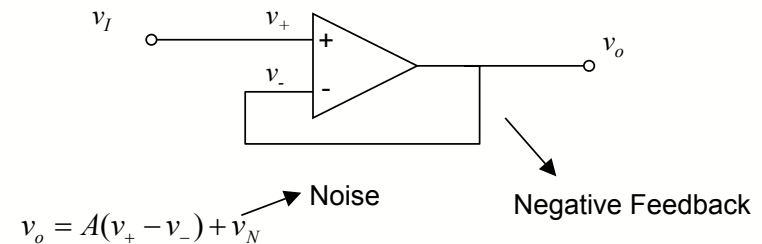
## Voltage Follower



Alternatively use virtual short principle for ideal op-amps:

$$v_+ = v_- \quad \longrightarrow \quad v_o = v_I$$

## Effect of Feedback on Noise



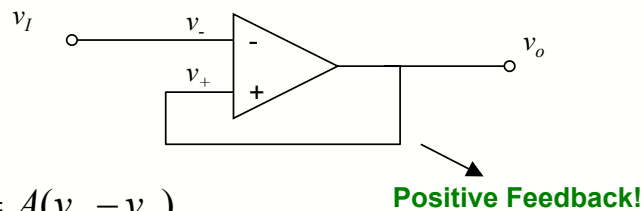
$$v_o = A(v_+ - v_-) + v_N$$

$$v_+ = v_I \quad v_- = v_o$$

$$v_o = A(v_I - v_o) + v_N = Av_I - Av_o + v_N$$

$$v_o = \frac{A}{A+1} v_I + \frac{v_N}{A+1} \quad \longrightarrow \quad \text{Noise reduced by a factor of } 1+A$$

## Positive Feedback



$$v_o = A(v_+ - v_-)$$

$$v_+ = v_o \quad v_- = v_I$$

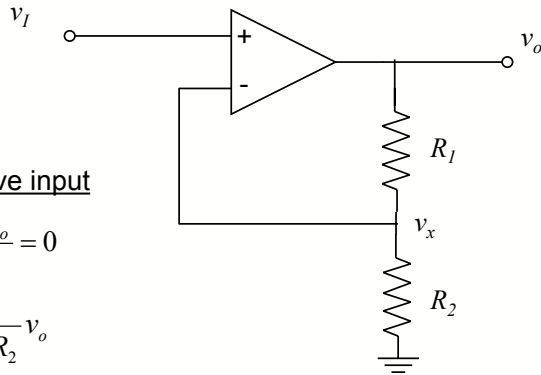
$$v_o = A(v_o - v_I) = Av_o - Av_I$$

$$\frac{v_o}{v_I} = \frac{-A}{1-A} \cong 1 \quad \longrightarrow \quad \text{Can be shown to be an unstable configuration.}$$

## Nodal Analysis of Op-Amp Circuits

1. Make use of the **virtual short/virtual open** principles.
2. Node voltages at the input are equal so one of them can be eliminated.
3. The currents at the input are zero and are involved in KCL equations at the **input** nodes.
4. The output current is **not** zero.
5. Make sure op-amp is in linear region so that virtual open/short principles can be applied (What happens when we have positive feedback?).

## Example: Non-Inverting Amp



KCL at negative input

$$\frac{v_x}{R_2} + \frac{v_x - v_o}{R_1} = 0$$

$$v_x = \frac{R_2}{R_1 + R_2} v_o$$

Virtual short

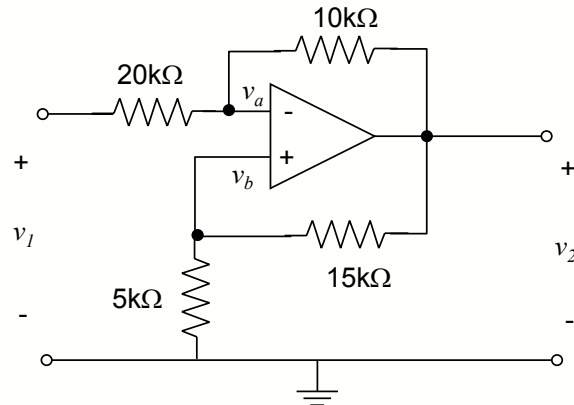
$$v_x = v_I \quad \longrightarrow \quad v_I = \frac{R_2}{R_1 + R_2} v_o$$

## Typical Uses of OpAmps

1. Buffer circuit.
2. Voltage scaling.
3. Analog Computers (solution of differential equations.). OpAmp circuits can be used to perform mathematical operations (addition, subtraction, integration, etc.).
4. Negative resistor (active component).
5. Active filters.

## Example: Inverting Amplifier

Textbook Exercise 4.8.3 (p.154)



KCL at negative input

$$\frac{v_a - v_1}{20k} + \frac{v_a - v_2}{10k} + i_- = 0$$

$\left. \begin{matrix} i_- = 0 \\ \text{(Virtual open principle)} \end{matrix} \right\}$

KCL at positive input

$$\frac{v_b}{5k} + \frac{v_b - v_2}{15k} + i_+ = 0$$

$\left. \begin{matrix} i_+ = 0 \\ \text{(Virtual open principle)} \end{matrix} \right\}$

$$v_a = v_b \quad \longrightarrow \quad \text{Virtual short principle}$$

## Example: Inverting Amplifier

---

First equation:  $\rightarrow \frac{v_a}{20} + \frac{v_a}{10} = \frac{v_1}{20} + \frac{v_2}{10} = \frac{3v_a}{20}$

Second equation:  $\rightarrow \frac{v_b}{5} + \frac{v_b}{15} = \frac{v_2}{15}$

$$4v_b = v_2$$

Third equation:  $\rightarrow v_a = v_b$

$\rightarrow \frac{v_1}{20} + \frac{v_2}{10} = \frac{3v_2}{80}$   $\rightarrow v_2 = -\frac{4}{5}v_1$

## Hints

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1. Apply KCL at the input of the op-amp and take advantage of the virtual open principle.
2. Never apply KCL at the output node of the op-amp unless you are asked to calculate its output current.
3. Never apply KCL at the reference node. Remember we do not show the currents into the power supplies of the op-amp.
4. Make use of the virtual short principle (the input voltages of the op-amp are equal).