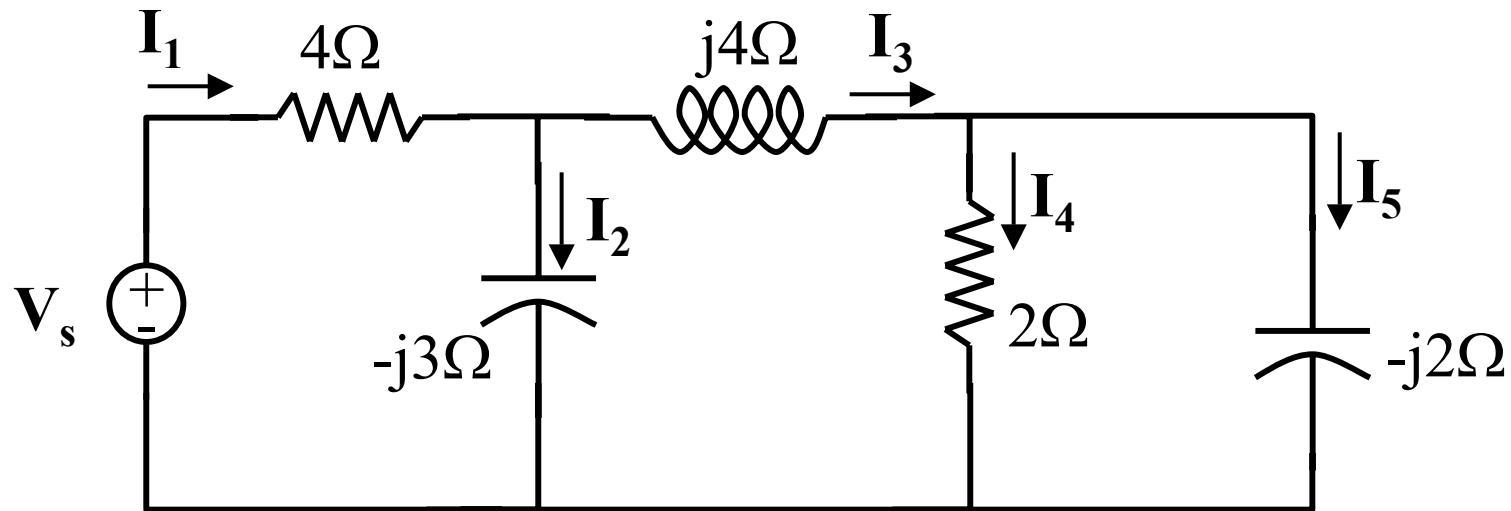


ECSE 210: Circuit Analysis

Lecture #13:

AC Steady State Analysis

Example 1: Linearity



Use linearity to determine the phasor \mathbf{I}_4 and the time domain current $i_4(t)$ given that:

$$v_s(t) = 12 \cos(377t + 30)V$$

→ The phasor for $v_s(t)$ is given by:

$$\mathbf{V}_s = 12 \angle 30^\circ V$$

Example 1: Linearity

Solve using linearity:

Assume: $I_4 = 1\angle 0$ → Solve for V_s

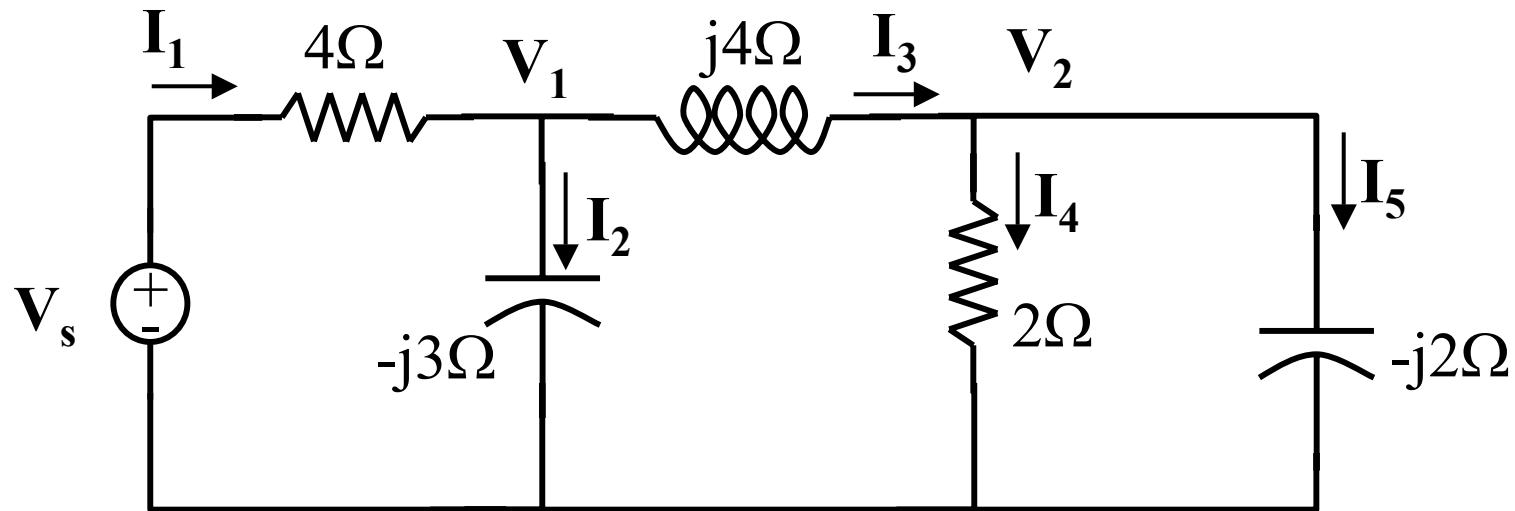
→ Using linearity calculate I_4 for the given value of V_s

For example:

- Assume that for $I_4=1A$ we can solve for $V_s=3V$.
- Then we could immediately state (using linearity) that if $V_s=1.5$ then $I_4 = (1.5/3)1 = 0.5A$

→ Apply above procedure to the circuit on previous slide.

Example 1: Linearity



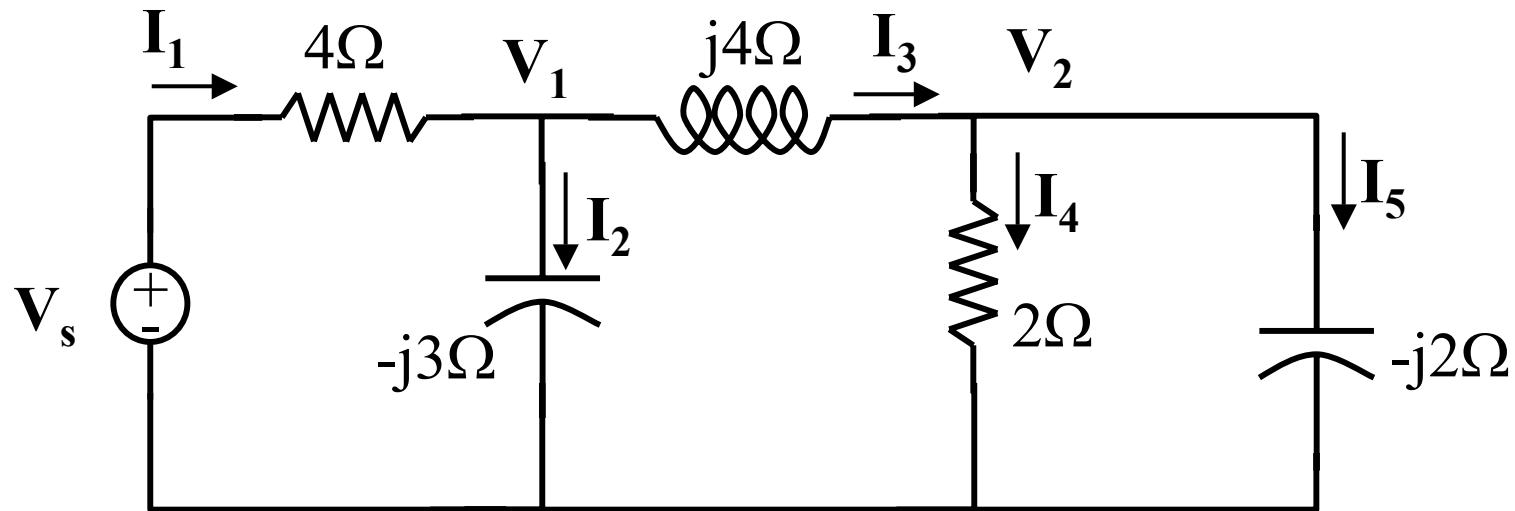
$$I_4 = 1 \angle 0^\circ A \quad \rightarrow \text{Calculate } V_s$$

$$I_4 = 1 \angle 0^\circ A \quad \rightarrow \quad V_2 = (1 \angle 0^\circ A)(2\Omega) = 2V$$

$$\rightarrow I_5 = \frac{2V}{-j2\Omega} = j$$

$$\rightarrow I_3 = I_4 + I_5 = 1 + j \quad A$$

Example 1: Linearity

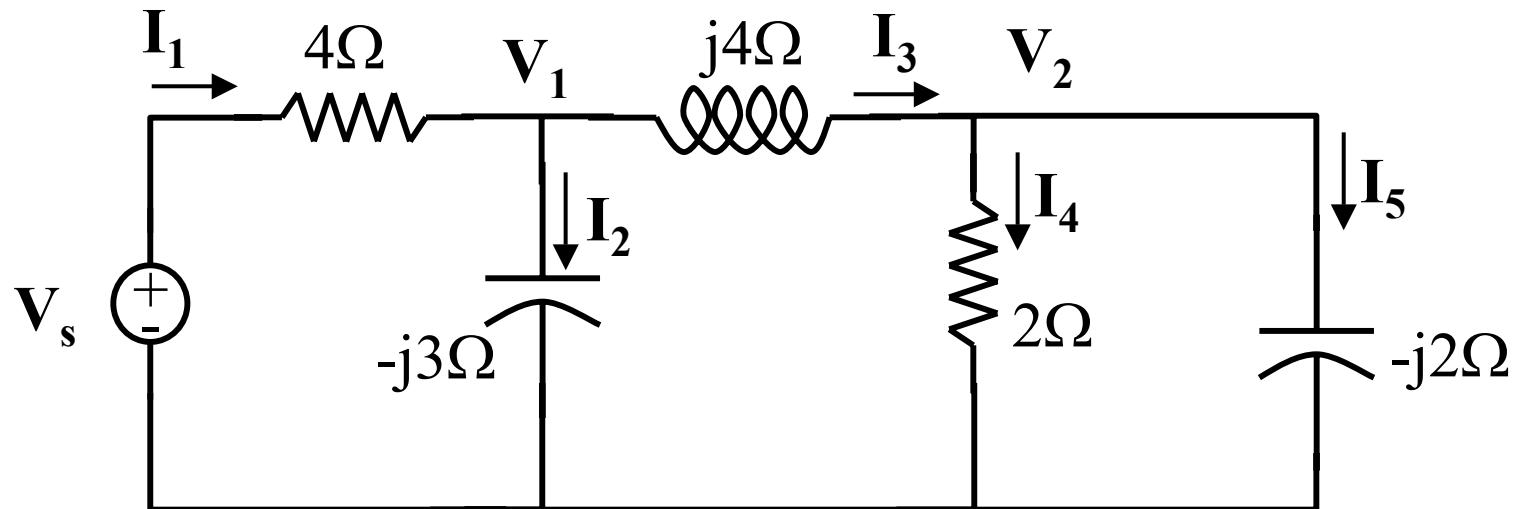


$$V_1 = I_3(j4\Omega) + V_2 = (1+j)(j4) + 2 = -2 + 4j \quad V$$

$$\rightarrow I_2 = \frac{-2 + 4j}{-j3} = \frac{-4}{3} - j \frac{2}{3}$$

$$\rightarrow I_1 = I_2 + I_3 = -\frac{4}{3} - j \frac{2}{3} + 1 + j = -\frac{1}{3} + j \frac{1}{3}$$

Example 1: Linearity



$$\rightarrow V_s = I_1(4\Omega) + V_1 = \left(-\frac{1}{3} + j\frac{1}{3}\right)(4\Omega) + (-2 + 4j)$$

$$\rightarrow V_s = -\frac{10}{3} + j\frac{16}{3} = 6.29 \angle 122^\circ \text{ V}$$

Example 1: Linearity

$$I_4 = 1\angle 0^\circ \rightarrow V_s = -\frac{10}{3} + j\frac{16}{3} = 6.29\angle 122^\circ$$

Using linearity:

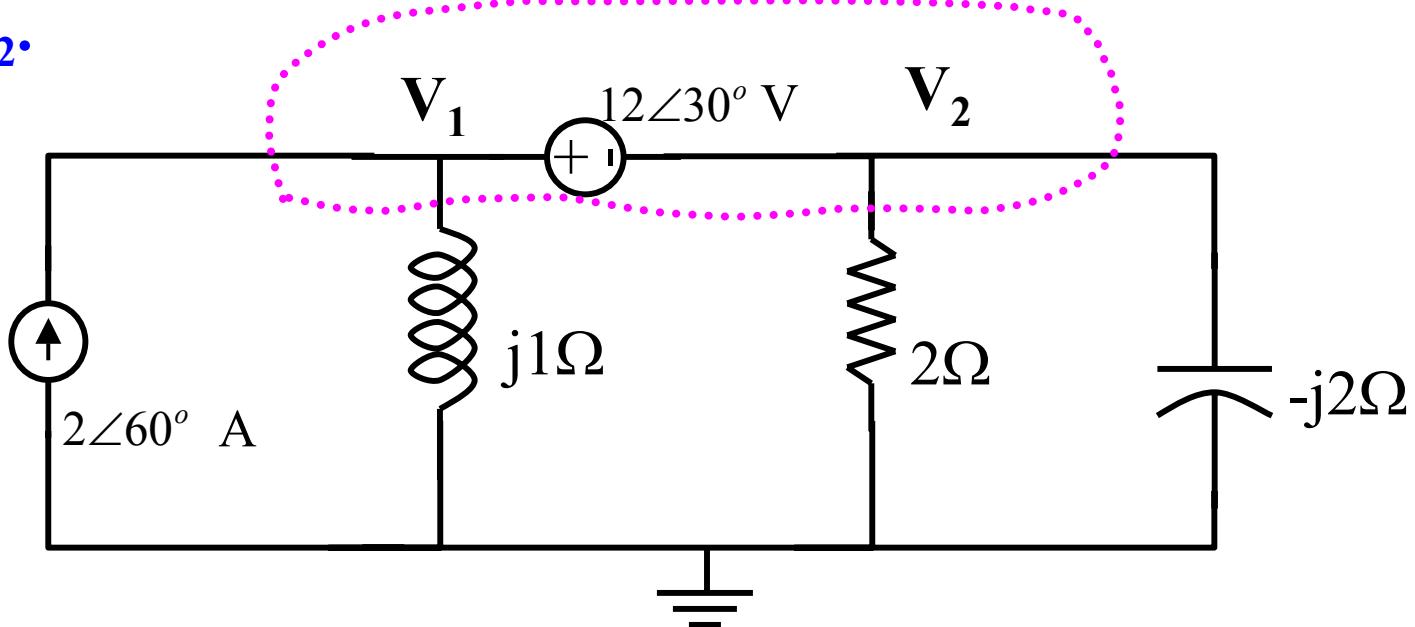
$$V_s = 12\angle 30^\circ \rightarrow I_4 = \frac{(12\angle 30^\circ)(1\angle 0^\circ)}{6.29\angle 122^\circ} = 1.91\angle -92^\circ$$

In the time domain:

$$i_4(t) = 1.91 \cos(377t - 92^\circ) \text{ A}$$

Example 2: Nodal Analysis

Find V_1 and V_2 .



KCL at supernode:

$$\frac{V_1}{j1} + \frac{V_2}{2} + \frac{V_2}{-j2} = 2\angle 60^\circ$$

For the supernode:

$$V_2 = V_1 - 12\angle 30^\circ$$

$$\left. \begin{aligned} \frac{V_1}{j1} + \frac{V_1 - 12\angle 30}{2} + \frac{V_1 - 12\angle 30}{-j2} &= 2\angle 60^\circ \end{aligned} \right\}$$

Example 2: Nodal Analysis

$$\frac{V_1}{j1} + \frac{V_1 - 12\angle 30}{2} + \frac{V_1 - 12\angle 30}{-j2} = 2\angle 60^\circ$$

$$V_1 \left(\frac{1}{j} + \frac{1}{2} + \frac{1}{-2j} \right) = 2\angle 60^\circ + \frac{12\angle 30^\circ}{2} - \frac{12\angle 30^\circ}{j2}$$

$$V_1 = 14.69\angle 117.14^\circ \text{ V}$$

Also

$$\begin{aligned} V_2 &= V_1 - 12\angle 30^\circ = 14.69\angle 117.14^\circ - 12\angle 30^\circ \\ &= -6.70 + j13.07 - (10.4 + j6) \\ &= -17.1 + j7.07 = 18.5\angle 157.54^\circ \text{ V} \end{aligned}$$

Example 3: Nodal Analysis 2

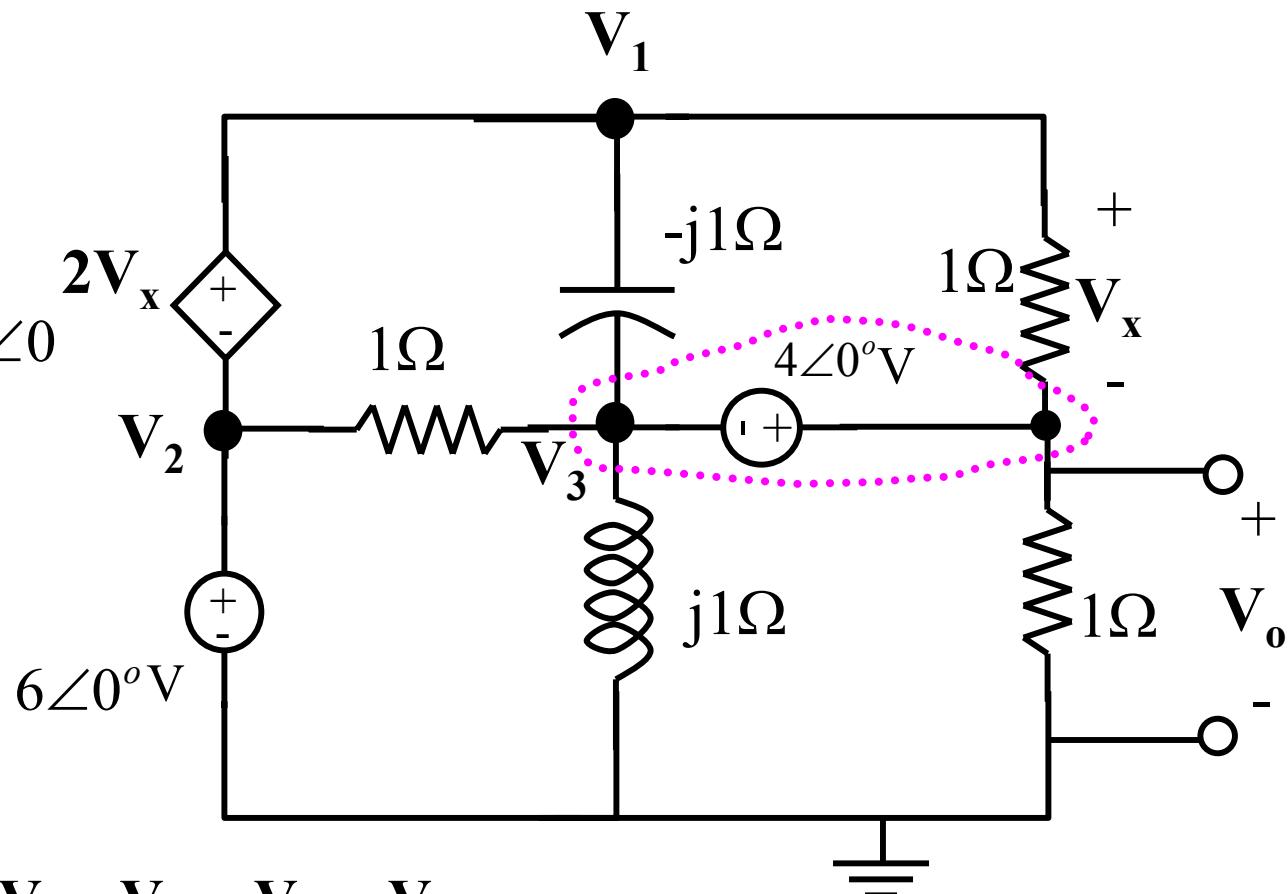
Find V_o .

$$V_2 = 6\angle 0$$

$$V_x = V_o - 6\angle 0$$

$$V_1 = V_o + V_x = 2V_o - 6\angle 0$$

$$V_3 = V_o - 4\angle 0$$



KCL at supernode:

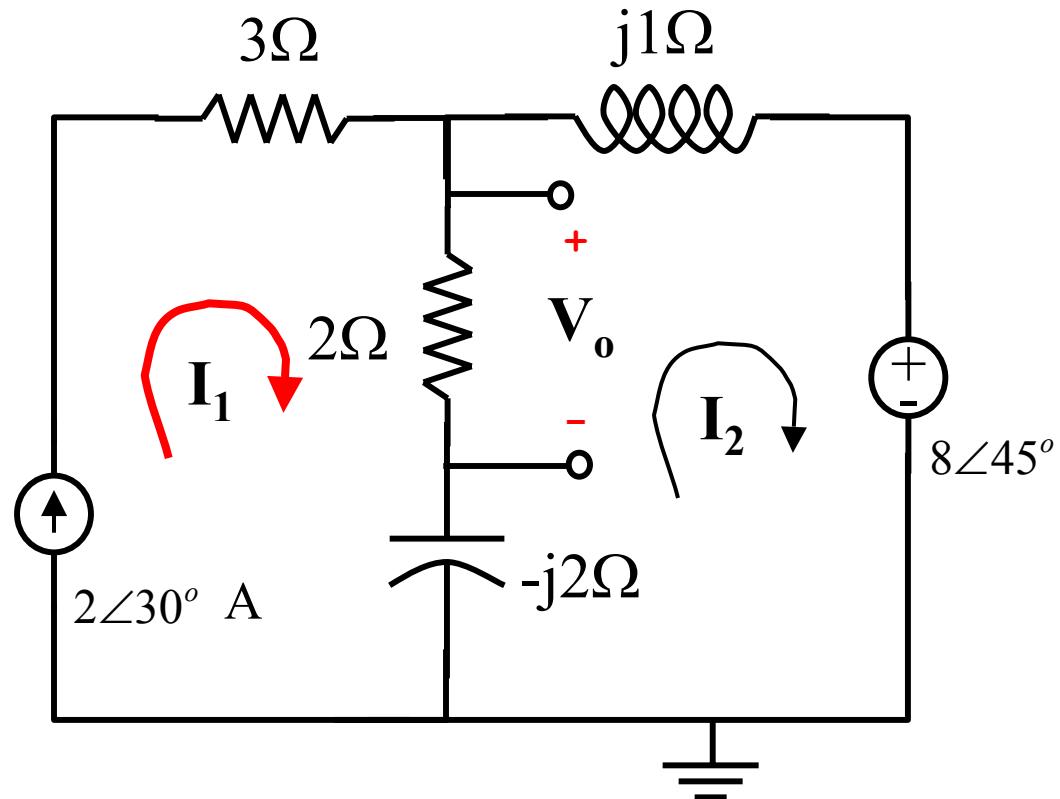
$$\frac{V_o - V_1}{1} + \frac{V_3 - V_1}{-j} + \frac{V_3 - V_2}{1} + \frac{V_3}{j} + \frac{V_o}{1} = 0$$

→ Substitute for V_1 , V_2 , V_3 : $V_o = 3.225\angle 7.125^\circ$ V

Example 4: Mesh Analysis

Find V_o .

$$I_1 = 2\angle 30^\circ$$



KVL

$$(I_2 - I_1)(2 - 2j) + I_2(j1) + 8\angle 45^\circ = 0$$

$$(-2\angle 30^\circ)(2 - 2j) + I_2(2 - 1j) + 8\angle 45^\circ = 0$$

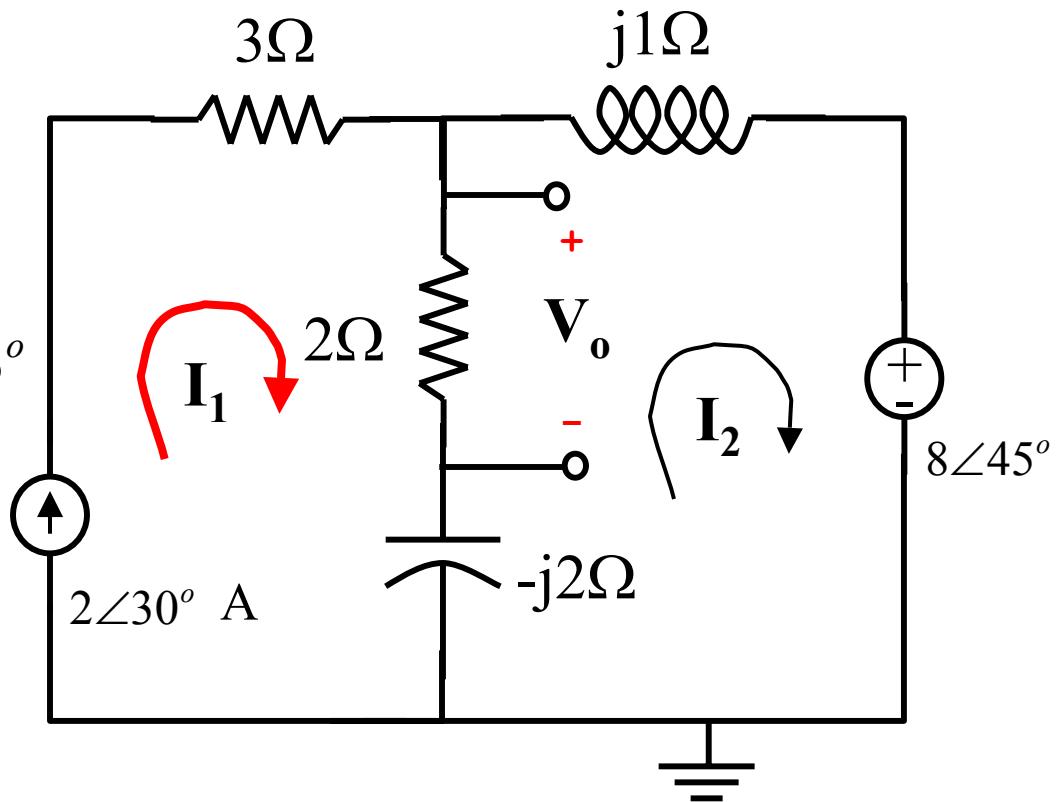
$$I_2 = \frac{(2\angle 30^\circ)(2 - 2j) - 8\angle 45^\circ}{2 - 1j} = 3.18\angle -64.96^\circ$$

Example 4: Mesh Analysis

$$\mathbf{I}_1 = 2\angle 30^\circ$$

$$\mathbf{I}_2 = 3.18\angle -64.96^\circ$$

$$\begin{aligned}\mathbf{I}_1 - \mathbf{I}_2 &= 2\angle 30^\circ - 3.18\angle -64.96^\circ \\ &= 0.38 + j3.88 \text{ A}\end{aligned}$$

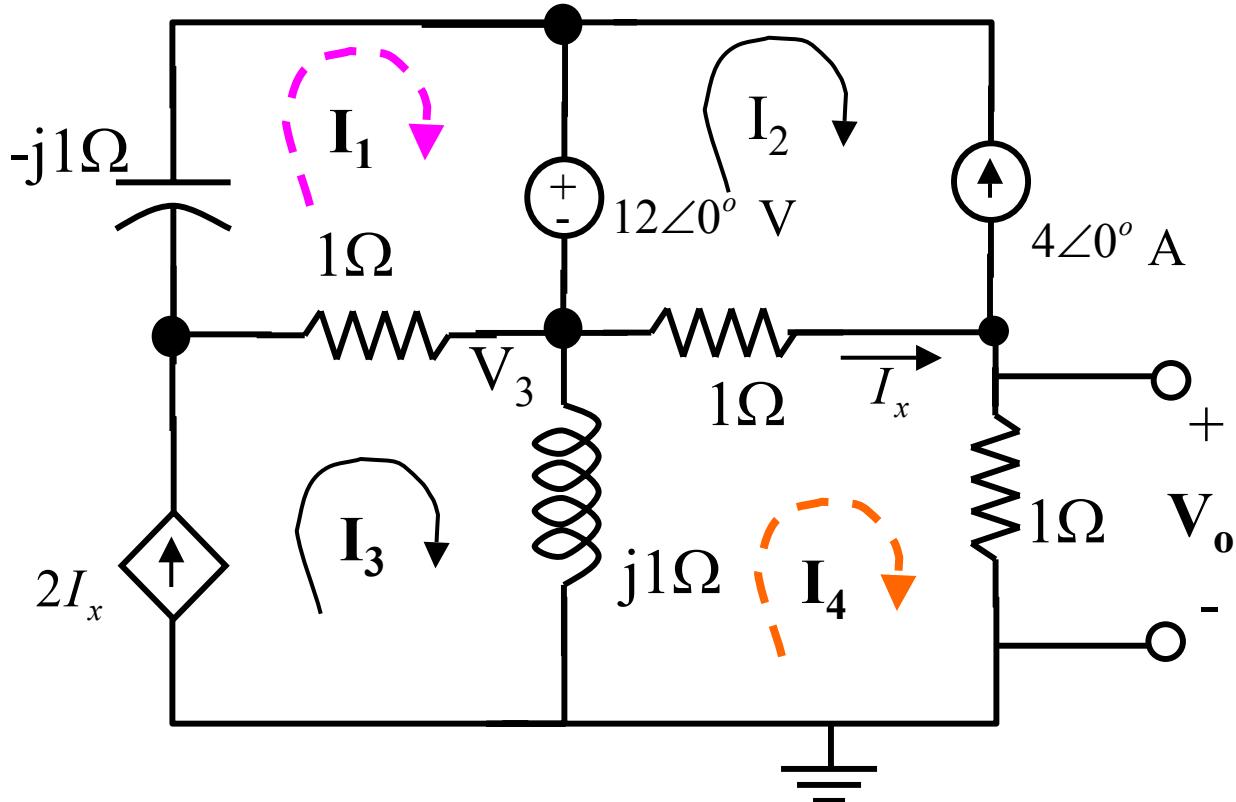


$$\mathbf{V}_o = 2(\mathbf{I}_1 - \mathbf{I}_2) = 0.76 + j7.76 \text{ V}$$

Example 5: Mesh Analysis 2

Objective:

Find I_4 then V_o .



Constraint Equations

$$I_2 = -4\angle 0$$

$$I_x = I_4 - I_2 = I_4 + 4$$

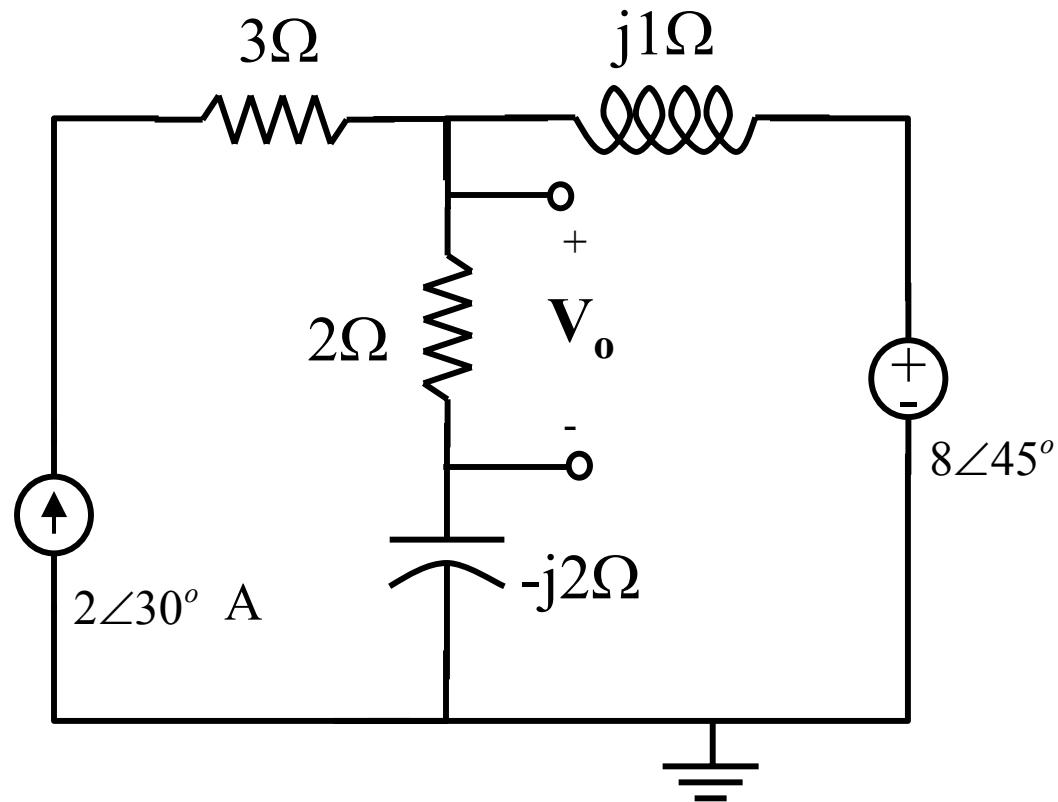
$$I_3 = 2I_x = 2(I_4 + 4)$$

Mesh Equations

$$\left. \begin{array}{l} -jI_1 + 1(I_1 - I_3) = -12\angle 0 \\ j(I_4 - I_3) + 1(I_4 - I_2) + 1I_4 = 0 \end{array} \right\} \quad \begin{array}{l} I_4 = -4\angle -36.87^\circ \text{ A} \\ V_o = -4\angle -36.87^\circ \text{ V} \end{array}$$

Example 6: Linear Superposition

Calculate V_o using superposition:



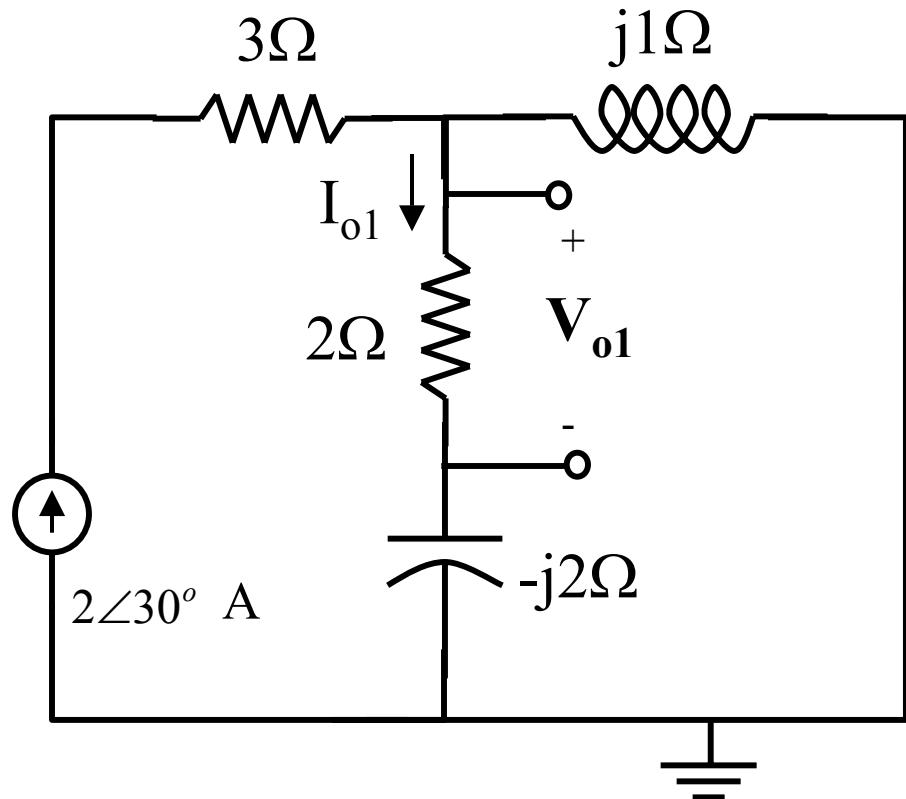
Example 6: Linear Superposition

Short circuit voltage source and calculate V_{o1} .

Current divider:

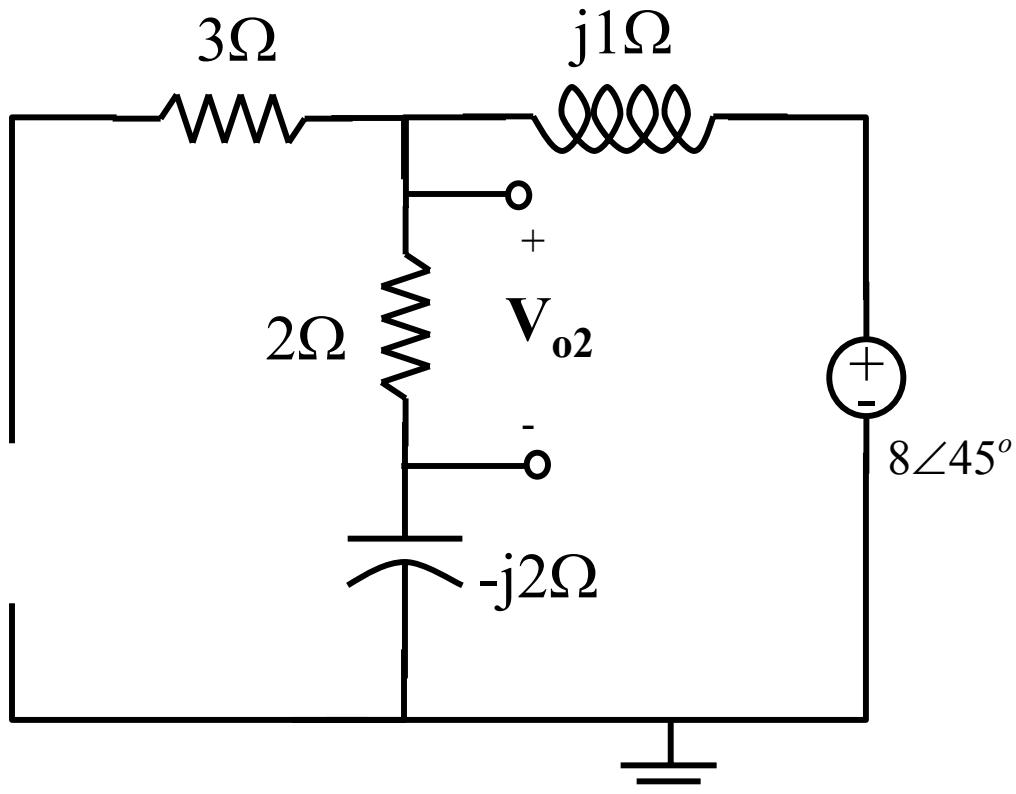
$$I_{o1} = \frac{j1\Omega}{2 - j2 + j1} (2\angle 30^\circ)$$

$$V_{o1} = \frac{j1}{2 - j2 + j1} (2\angle 30^\circ)(2) = 1.79\angle 146.57^\circ$$



Example 6: Linear Superposition

Open circuit current source and calculate V_{o2} .

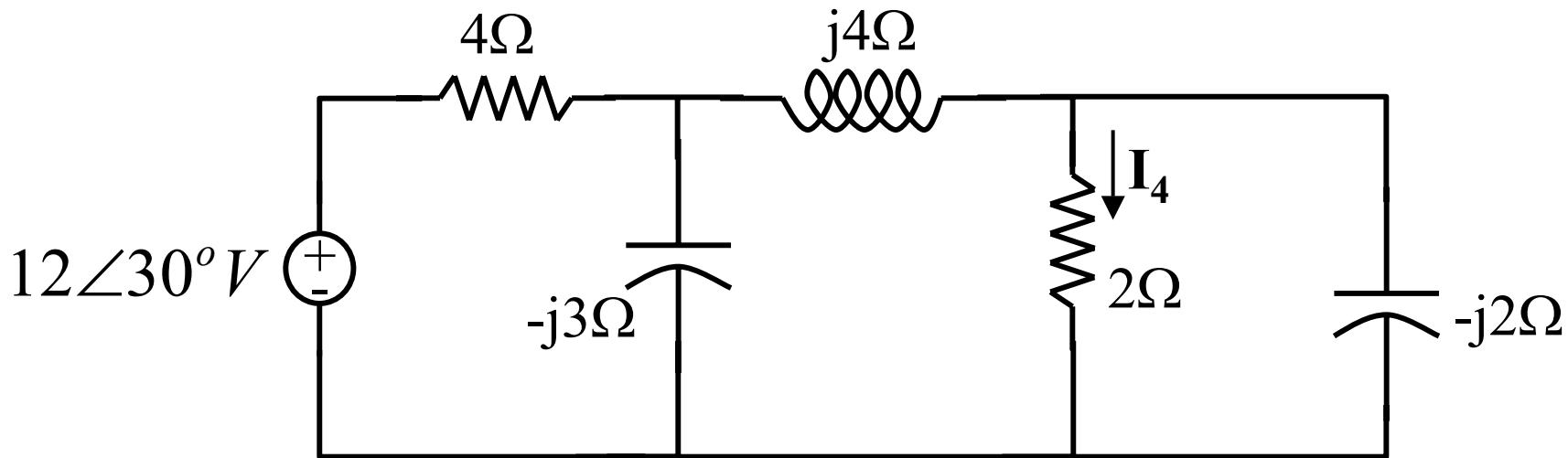


Voltage divider:

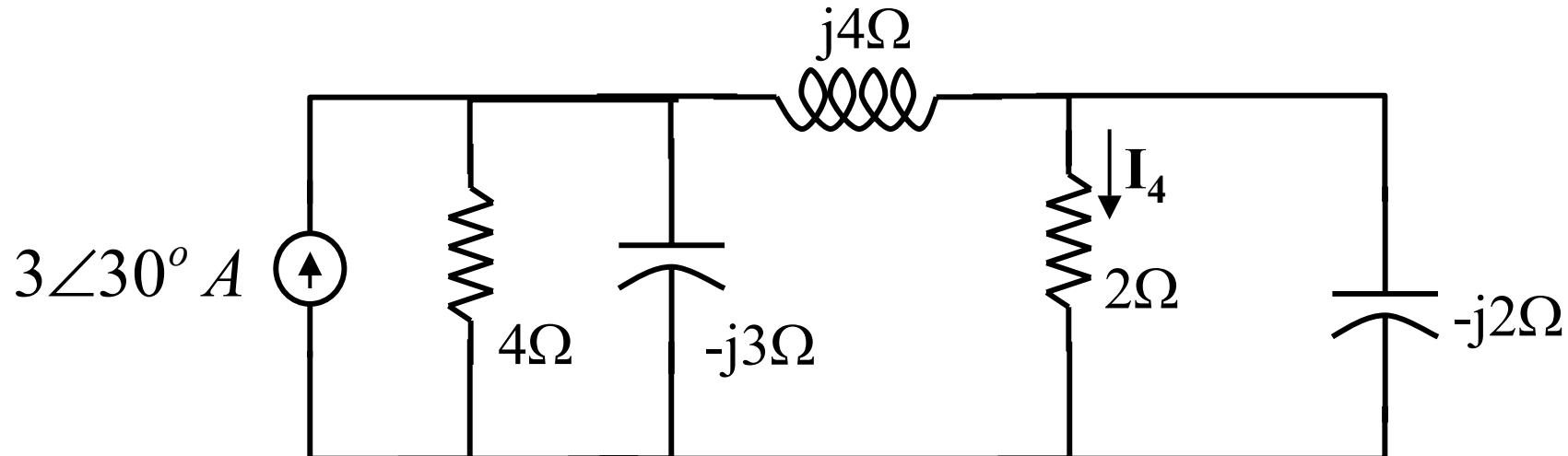
$$V_{o2} = \frac{2}{-j2 + 2 + j1} 8\angle 45^\circ = 7.14\angle 71.57^\circ$$

By superposition: $V_o = V_{o1} + V_{o2} = 1.79\angle 146.57^\circ + 7.14\angle 71.57^\circ$
 $= 0.76 + j7.76 \text{ V}$

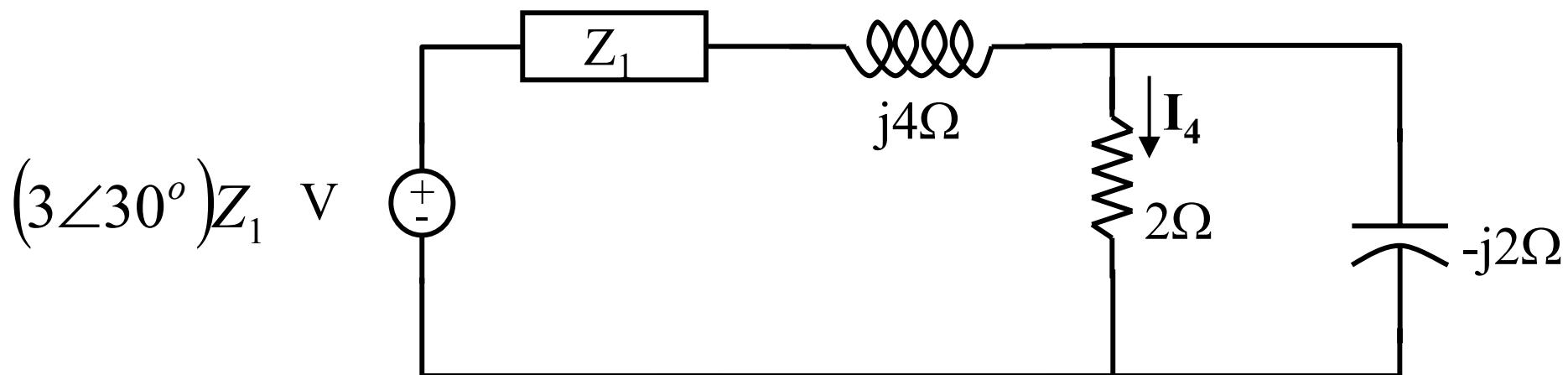
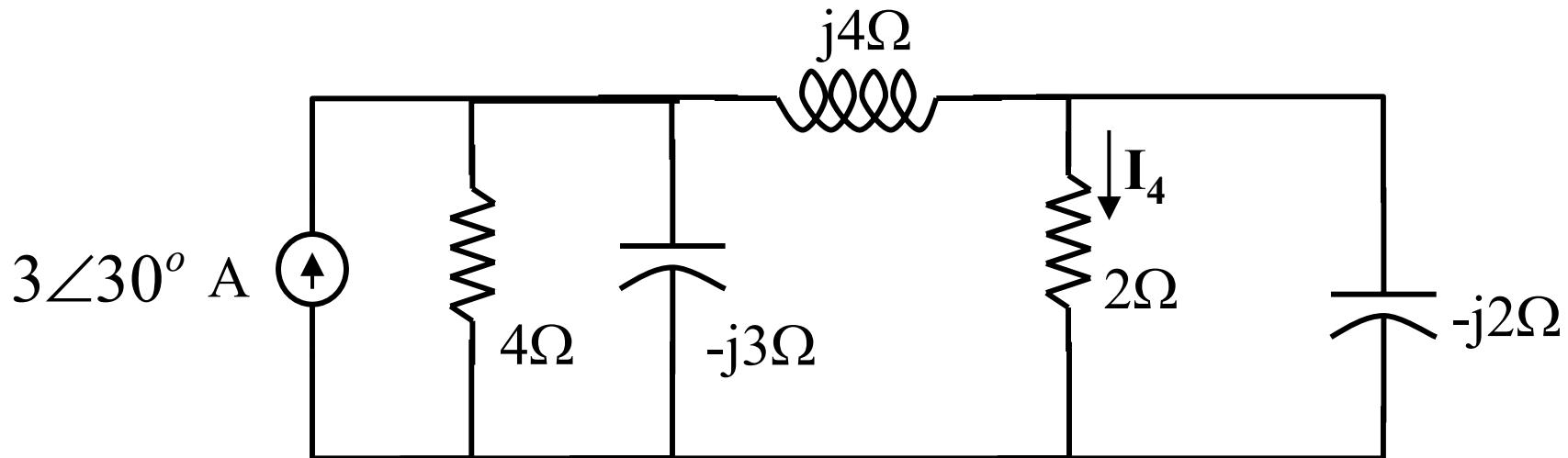
Example 7: Source Transformation



Use source transformation to determine the phasor I_4 .

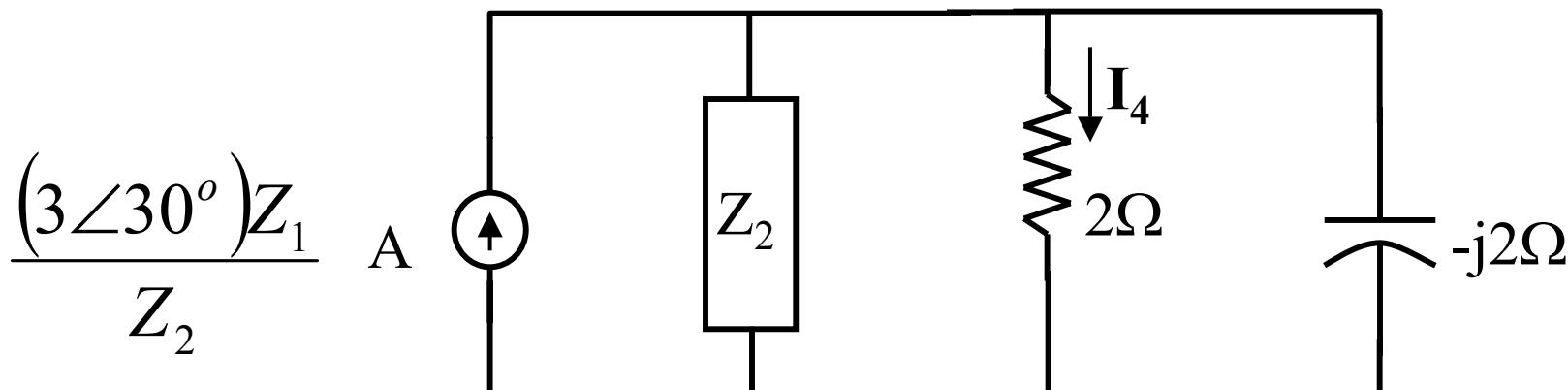
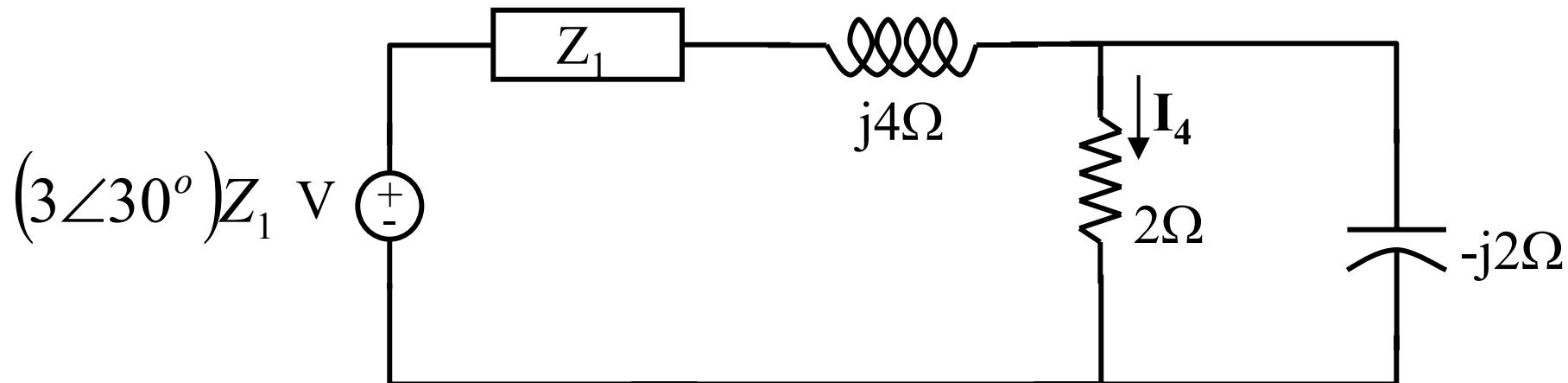


Example 7: Source Transformation



$$Z_1 = \frac{(4)(-j3)}{4 - j3} = \frac{-j12}{4 - j3}$$

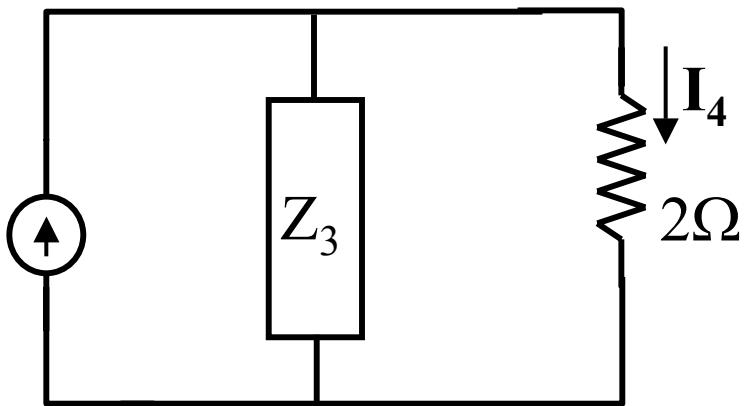
Example 7: Source Transformation



$$Z_2 = Z_1 + j4 = \frac{12 + j4}{4 - j3} \quad \Omega$$

Example 7: Source Transformation

$$\frac{(3\angle 30^\circ)Z_1}{Z_2}$$



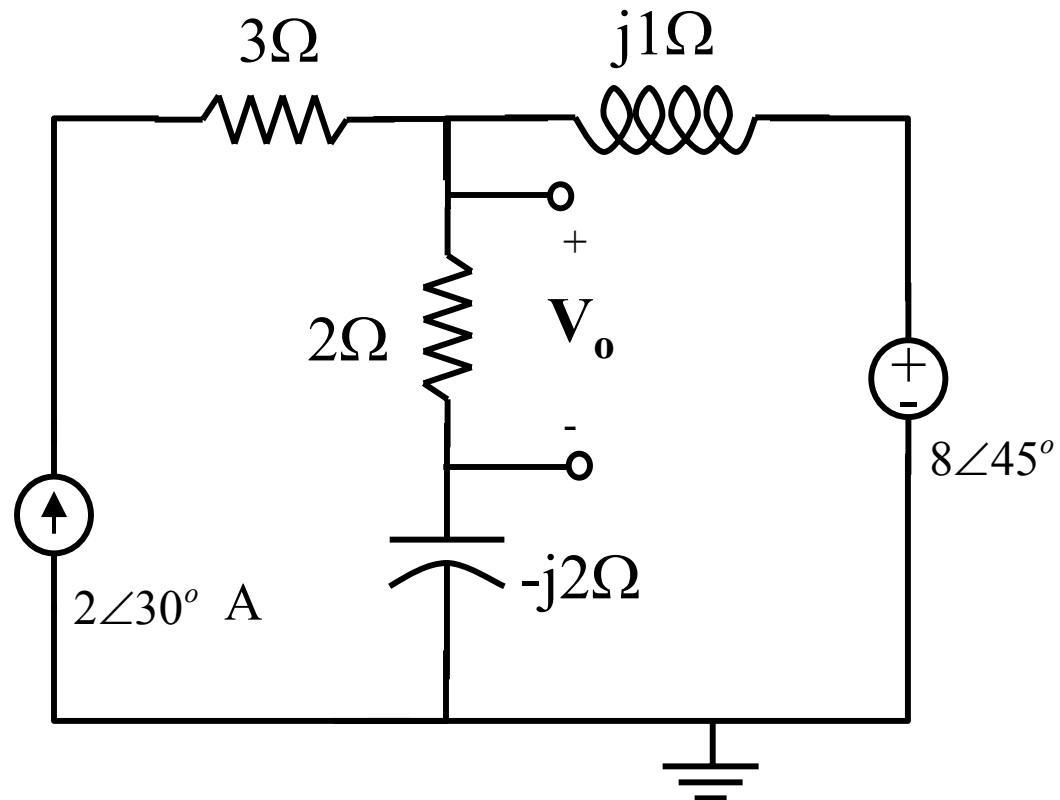
$$Z_3 = \frac{(-j2)(Z_2)}{-j2 + Z_2}$$
$$= \frac{4 - j12}{3 - j2} \quad \Omega$$

$$\frac{(3\angle 30^\circ)Z_1}{Z_2} = 2.85\angle -78.43^\circ \text{ A}$$

$$I_4 = \frac{\left(\frac{4 - j12}{3 - j2}\right)}{2 + \left(\frac{4 - j12}{3 - j2}\right)} 2.85\angle -78.43^\circ = 1.91\angle -78.43^\circ \text{ A}$$

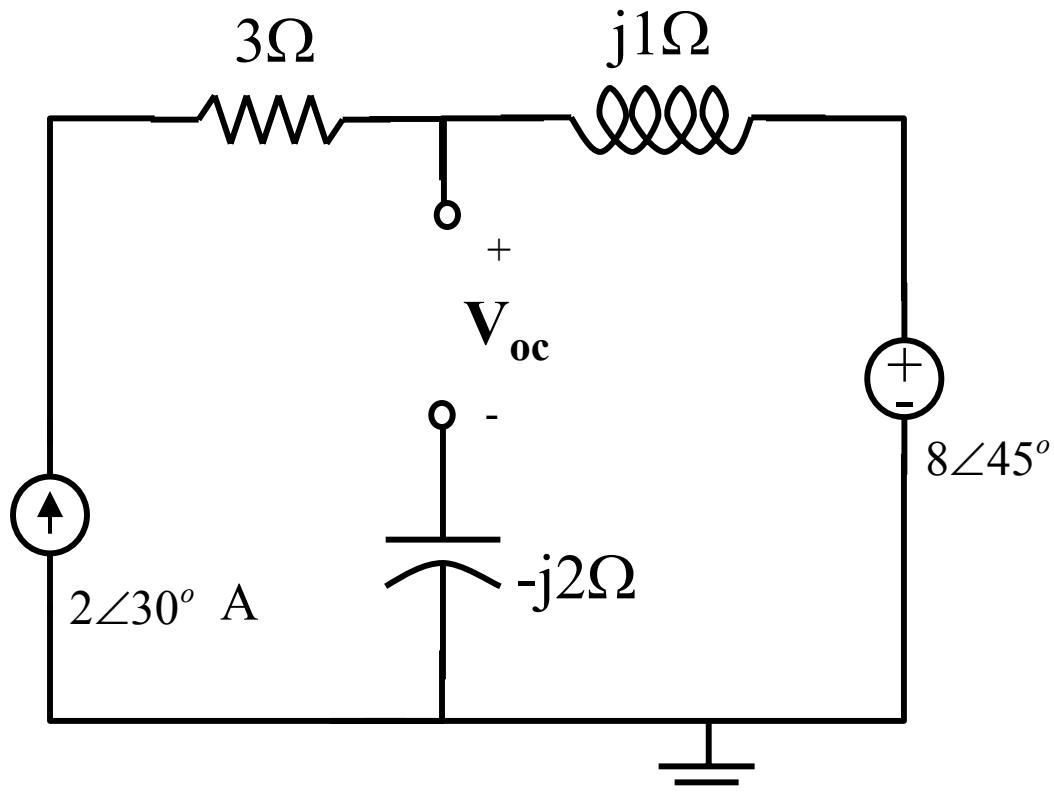
Example 8: Thevenin's Theorem

Calculate V_o using
Thevenin's Theorem.



Example 8: Thevenin's Theorem

Calculate V_o using
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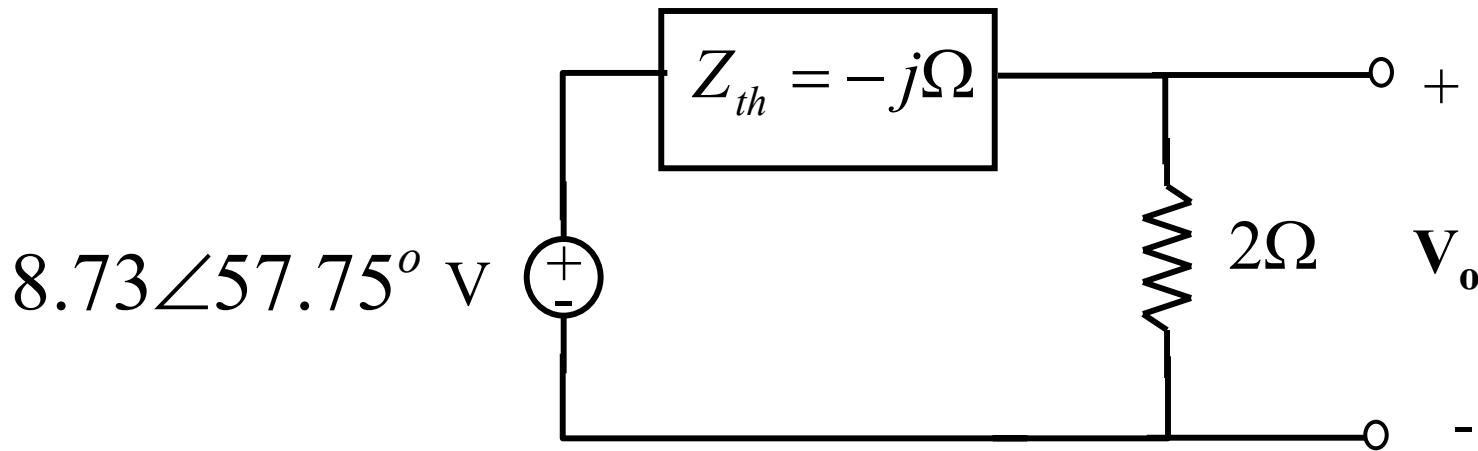


$$V_{oc} = (2\angle30^\circ)(j1) + 8\angle45^\circ = 8.73\angle57.75^\circ \text{ V}$$

$$Z_{th} = -j2 + j1 = -j\Omega$$

$$V_{th} = V_{oc} = 8.73\angle57.75^\circ \text{ V}$$

Example 8: Thevenin's Theorem



$$V_o = \frac{2}{2-j} 8.73\angle 57.75^\circ = 0.76 + j7.76 \quad \text{V}$$

Summary

1. All the methods derived for solving *dc steady-state* circuits are also valid for *ac steady-state* circuit analysis in the frequency domain. These methods include:
 - Nodal analysis.
 - Mesh and loop analysis.
 - Superposition.
 - Source transformation.
 - Thevenin and Norton's theorems
2. The principle of linearity is valid for the analysis of *ac steady-state* circuit analysis in the frequency domain.